



THE UNIVERSITY OF  
**AUCKLAND**  
Te Whare Wānanga o Tāmaki Makaurau  
NEW ZEALAND

**Electeng 311**

# **Electronics Systems Design**

## **MOSFETs**

Seho Kim

# Contents

- What is a switch?
- Why MOSFETs?
- MOSFETs overview
- Different MOSFET types
- Characteristic curves
- Breakdown voltage
- Conduction loss
- Switching loss
- Safe operating area
- MOSFET drivers

# Learning outcomes

- Understand the operation of an ideal switch.
- Recognise different MOSFET types.
- Understand and be able to explain how MOSFETs turn on and off.
- Understand a MOSFET body diode.
- Able to explain different MOSFET operation regions.
- Calculate simple power loss in a MOSFET and understand absolute temperature ratings.
- Able to interpret a MOSFET datasheet for application requirements.
- Understand the need for a MOSFET driver.



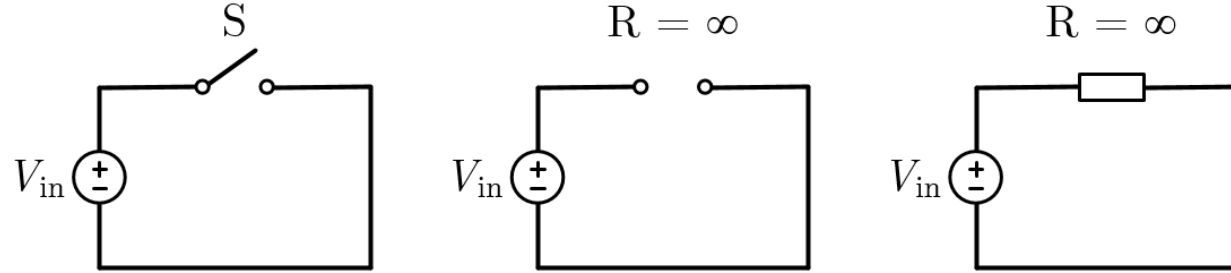
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# Switches

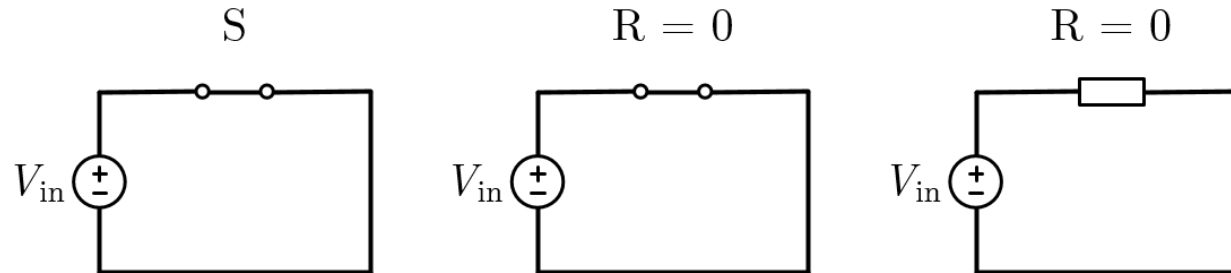
What is a switch?

# Ideal switch

- Electrical engineering relies on the ability to control the flow of current.
- A switch can ‘turn on’ to let the current flow.
- A switch can ‘turn off’ to stop the current flow.



- An opened ideal switch acts as an infinite impedance.



- A closed ideal switch acts as a zero impedance.

# Different types of switches

- Switches are used everywhere from light switches to microcontrollers.
- Oldest forms of switches were activated physically.
- Relays energise an electro-magnet to activate.
- Vacuum tubes were regularly used in early to mid 1900s.
- Large, expensive or short lifetime.



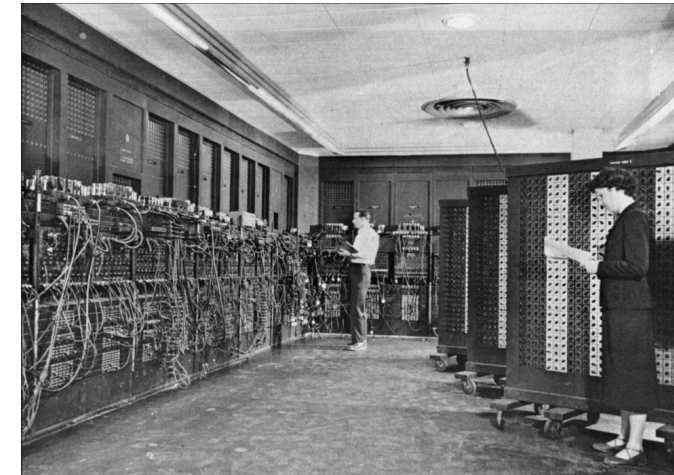
• Blade switch [1]



• Relay [2]



• Vacuum tube [3]



- ENIAC, the first fully electronic general-purpose programmable computer [4]

[1] <https://electronics.stackexchange.com/questions/199389/knife-blade-on-off-switch-for-an-a-c-circuit>

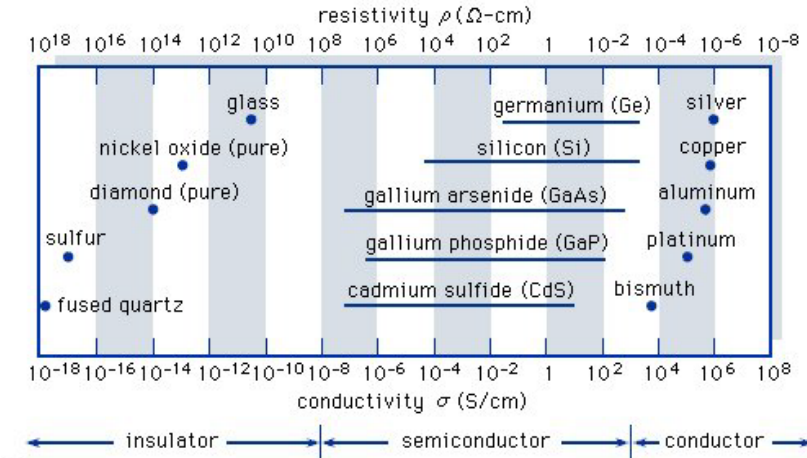
[2] [https://www.beta.com.tw/products\\_detail/54.htm](https://www.beta.com.tw/products_detail/54.htm)

[3] <https://www.tubesandmore.com/products/vacuum-tube-7025-wa-tube-amp-doctor-high-grade-premium-selected>

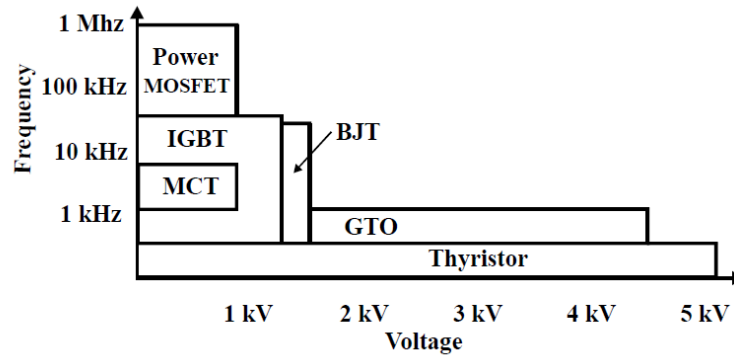
[4] M. Guarnieri, "The Age of Vacuum Tubes: Merging with Digital Computing [Historical]," in *IEEE Industrial Electronics Magazine*, vol. 6, no. 3, pp. 52-55, Sept. 2012, doi: 10.1109/MIE.2012.2207830.

# Semiconductors

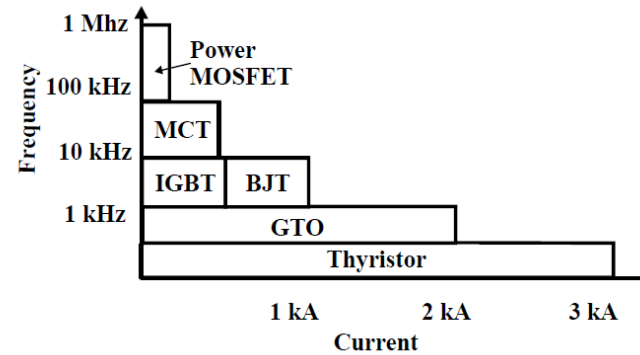
- The search for cheaper, smaller, efficient switches continued in mid-1900s.
- Semiconductors have conductivity somewhere between a conductor (e.g. copper) and an insulator (e.g. glass).
- Semiconductor switches are considered to be solid-state with no moving parts.
- Different semiconductor devices are suitable for different switching applications – breakdown voltage, switching time and efficiency.
- Bipolar junction transistors quite popular several decades ago.



- Conductivity of semiconductors [5]



- Power semiconductor operating regions [6]



- A piece of silicon [7]



- A transistor [8]

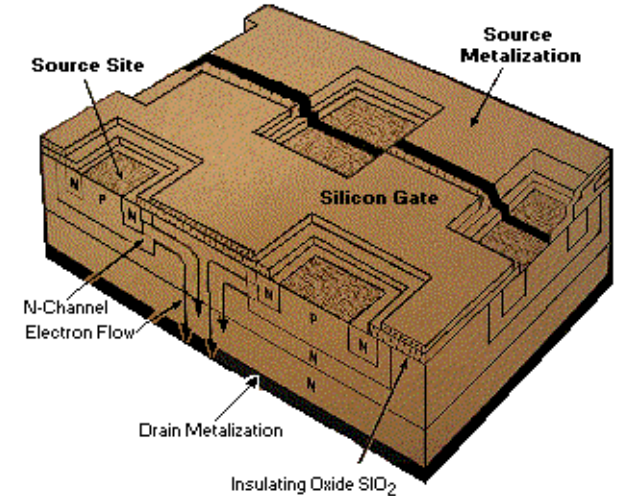
# MOSFETs

## Overview

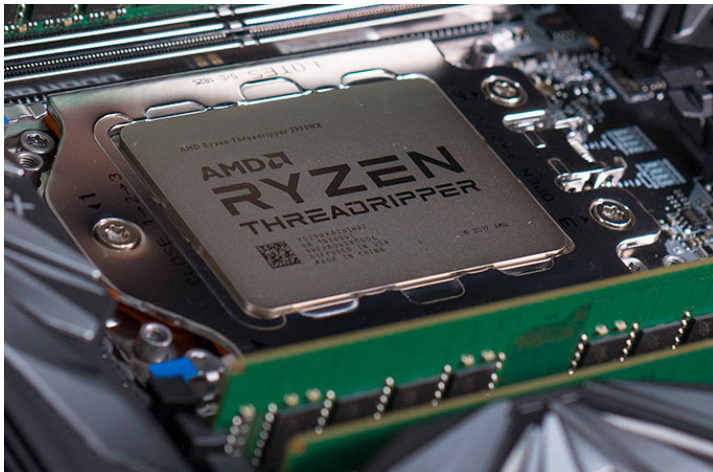


# MOSFETs – Metal-Oxide-Semiconductor Field-Effect Transistor

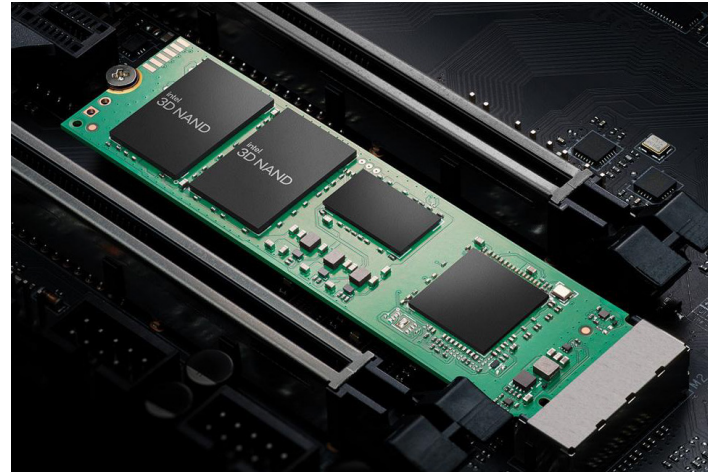
- Commonly used in electronics from CPUs to inverters.
- Micro metres (um) long. Most MOSFET devices are arrays of tiny MOSFET cells.
- Fast switching.
- Easily controllable.
- Relatively high voltage and current ratings.



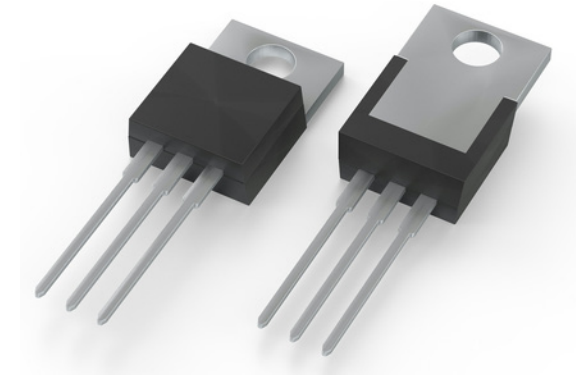
- Array of MOSFET cells [9]



- CPU [10]



- NAND flash drive [11]

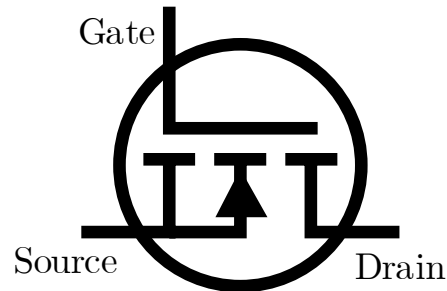


- Off the shelf MOSFETs [12]

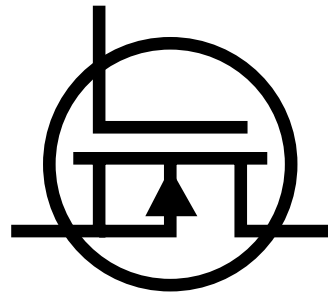
# MOSFETs – different types

- Arrow pointing up for N-type. Down for P-type.
- N-type enhancement is often used for switching applications.

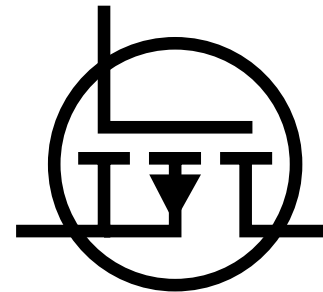
N-type enhancement



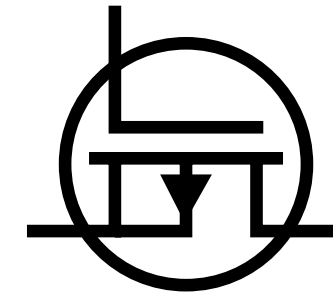
N-type depletion



P-type enhancement



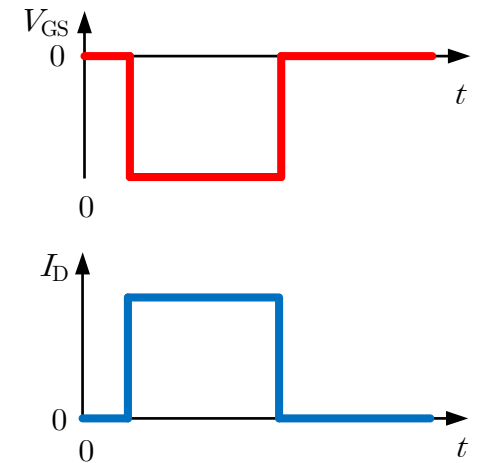
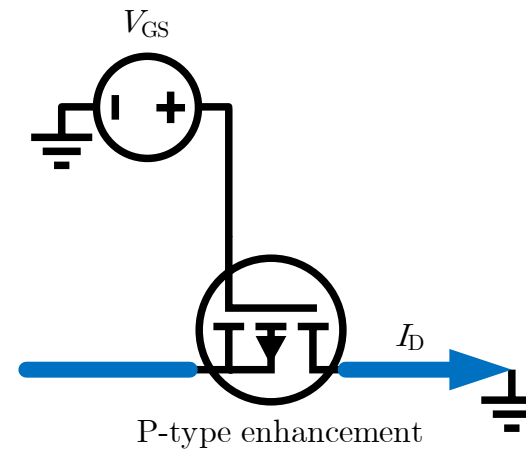
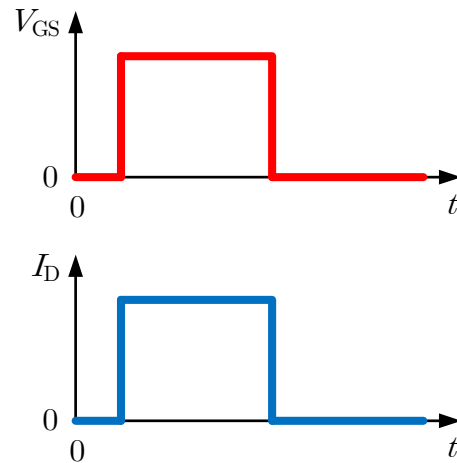
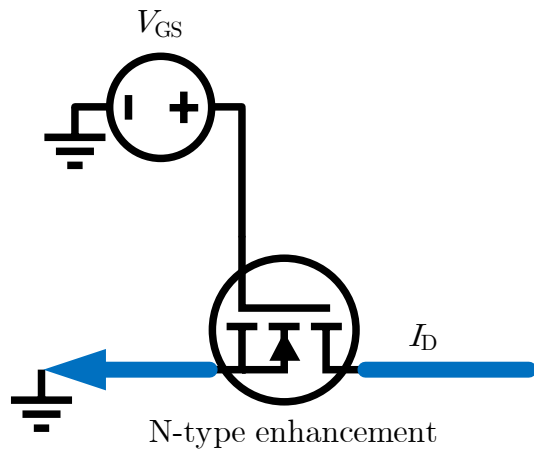
P-type depletion



MOSFET type	Positive $V_{GS}$	$V_{GS} = 0V$	Negative $V_{GS}$
N-type enhancement	ON	OFF	OFF
N-type depletion	ON	ON	OFF
P-type enhancement	OFF	OFF	ON
P-type depletion	OFF	ON	ON

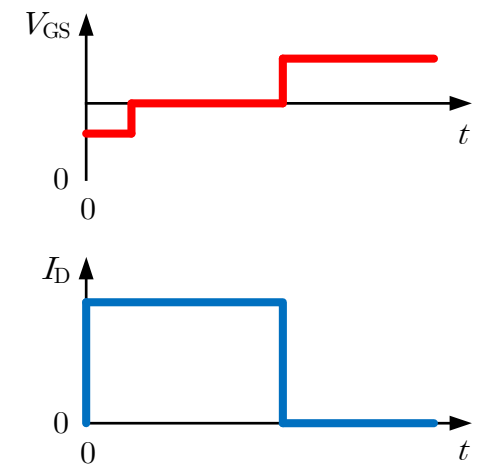
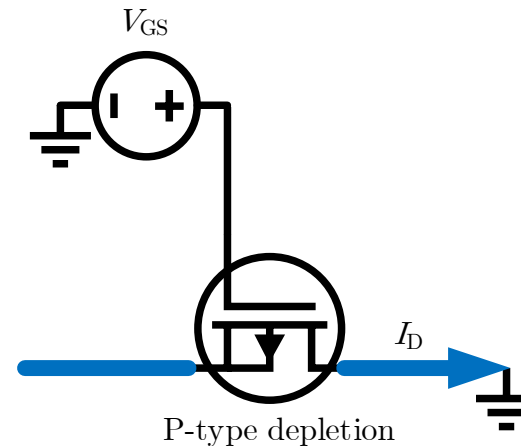
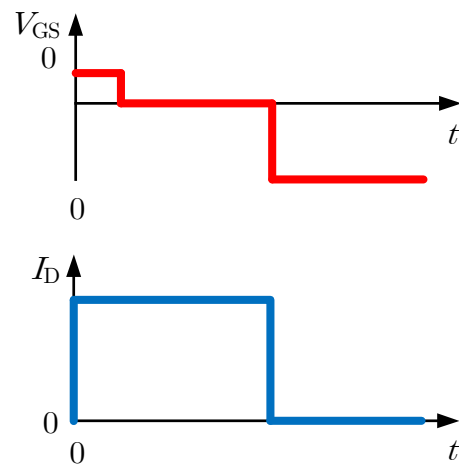
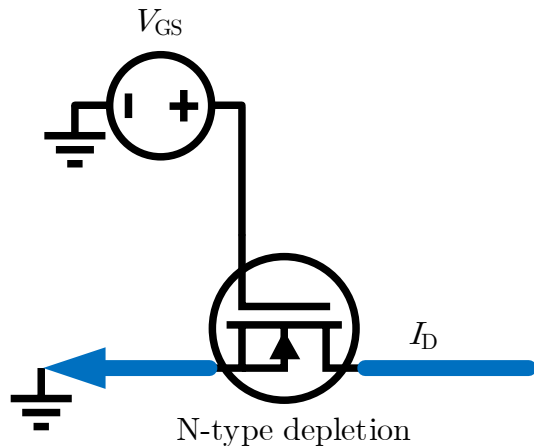
# MOSFETs – enhancement mode

- Two modes of MOSFETs: enhancement and depletion modes.
- Enhancement is the most common.
- Enhancement MOSFETs are nominally turned off. (Naturally turned off when no voltage is applied to the gate)
- N-type enhancement mode **turns on** with **positive voltage** at the gate.
- P-type enhancement mode **turns on** with **negative voltage** at the gate.



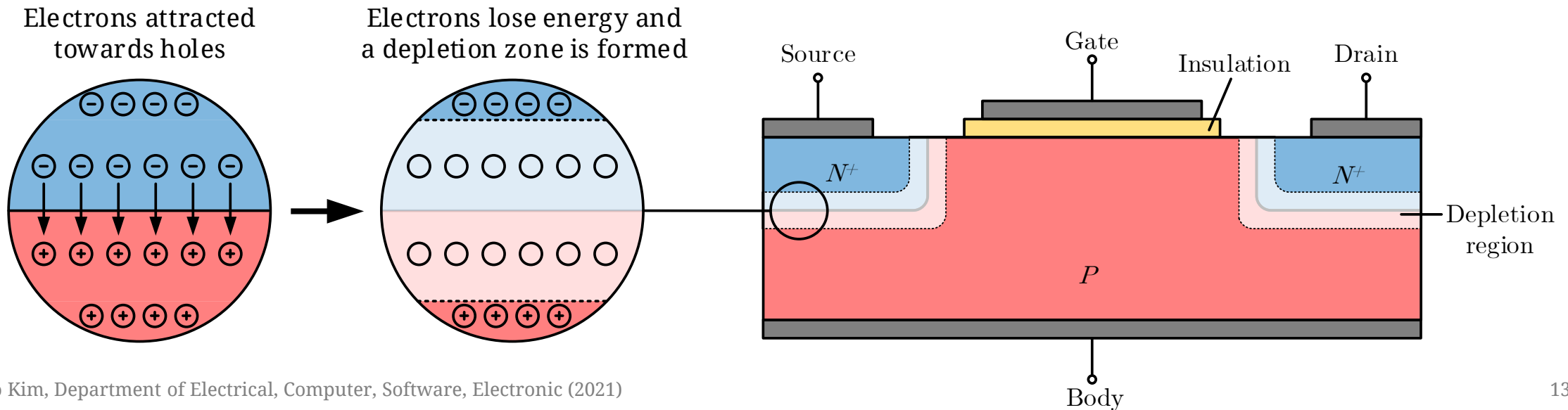
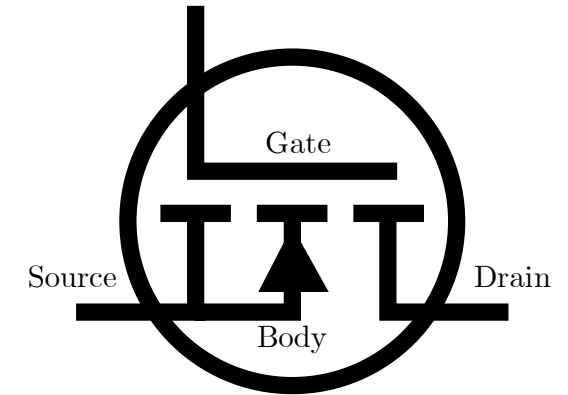
# MOSFETs – depletion mode

- Depletion MOSFETs are nominally turned on.
- Note that depletion mode MOSFETs can also go into enhancement mode.
- A depletion MOSFET turns on if no voltage applied at the gate or enhancement condition is met.
- N-type depletion mode **turns on** with **positive voltage or no voltage** at the gate.
- P-type depletion mode **turns on** with **negative voltage or no voltage** at the gate.



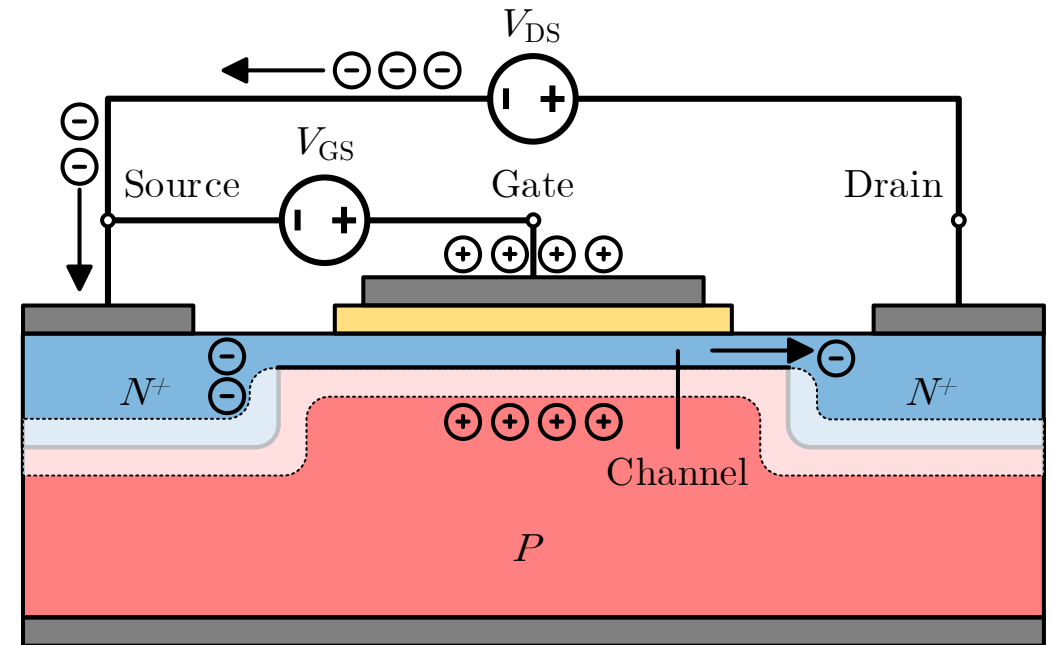
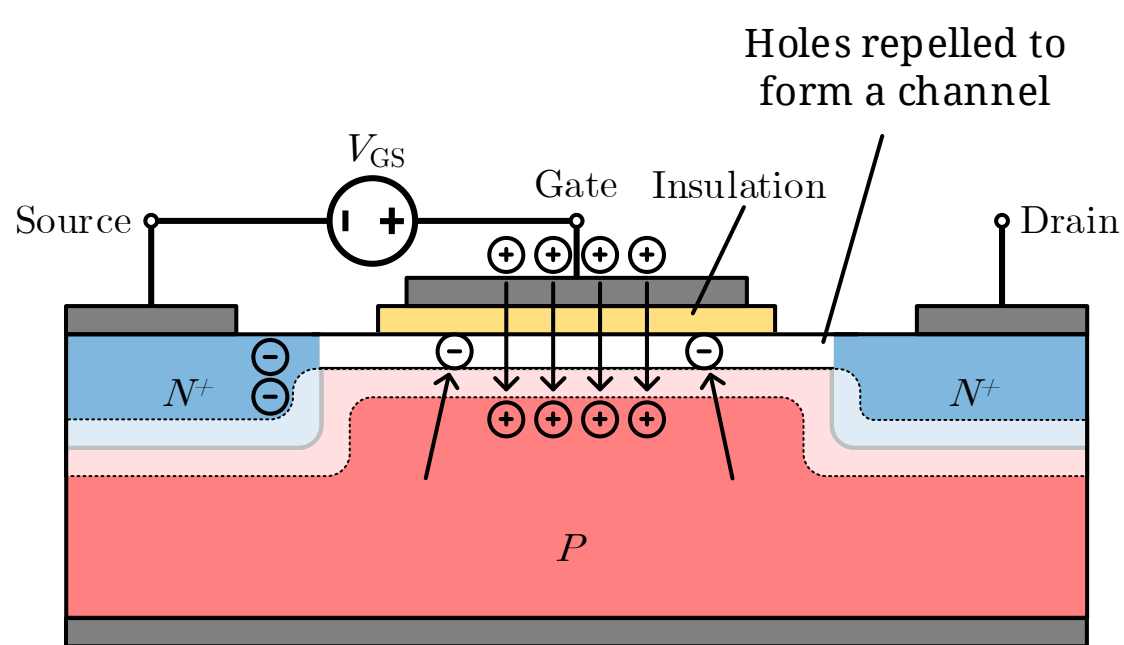
# MOSFETs – N-type enhancement structure

- N-type enhancement MOSFET has two N regions at the source and drain with a P region in between.
- P-N junctions: lots of electrons on one side and lots of holes on the other side.
- Electrons are naturally attracted to holes.
- Once electrons reach the holes in a P-N junction, their energy to move is 'depleted'.
- This forms depletion zones in the middle of P-N junctions.
- At rest, no electron can flow from the source to drain of the MOSFET.



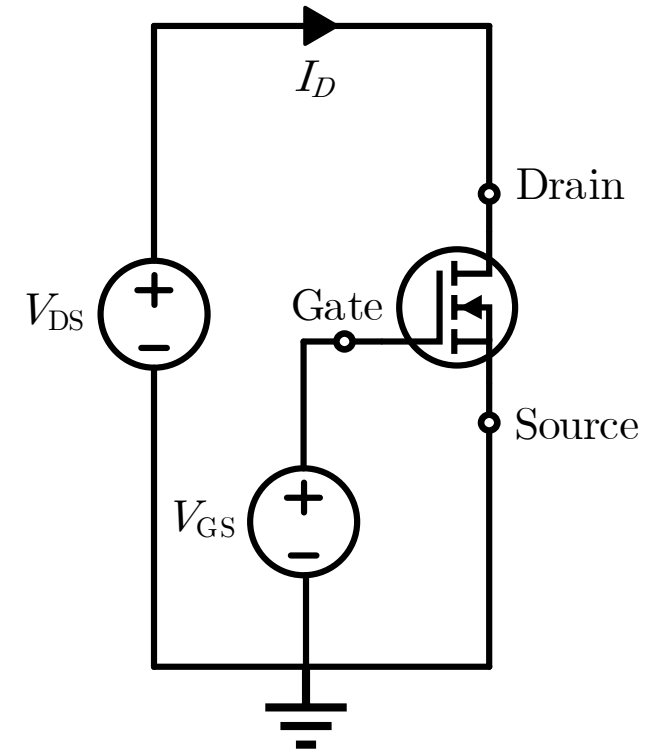
# MOSFETs – N-type enhancement structure

- Positive voltage applied to the gate ‘repels’ the holes near the gate, while electrons move towards the gate.
- This forms a channel between the source and the drain.
- When an external voltage source is applied, electrons can freely flow from the source to the drain through the channel.
- Note that **conventional current flows from the drain to the source**.



# MOSFETs – overview

- Today, MOSFETs are predominantly used to process information and power.
- N-type and P-type MOSFETs.
- Enhancement and depletion modes.
- Electrons always move towards holes.
- P-N junctions and depletion zone.
- N-type enhancement most commonly used for switching purposes.
- Positive voltage to the gate to turn on the N-type MOSFET.



- An example N-type enhancement MOSFET circuit

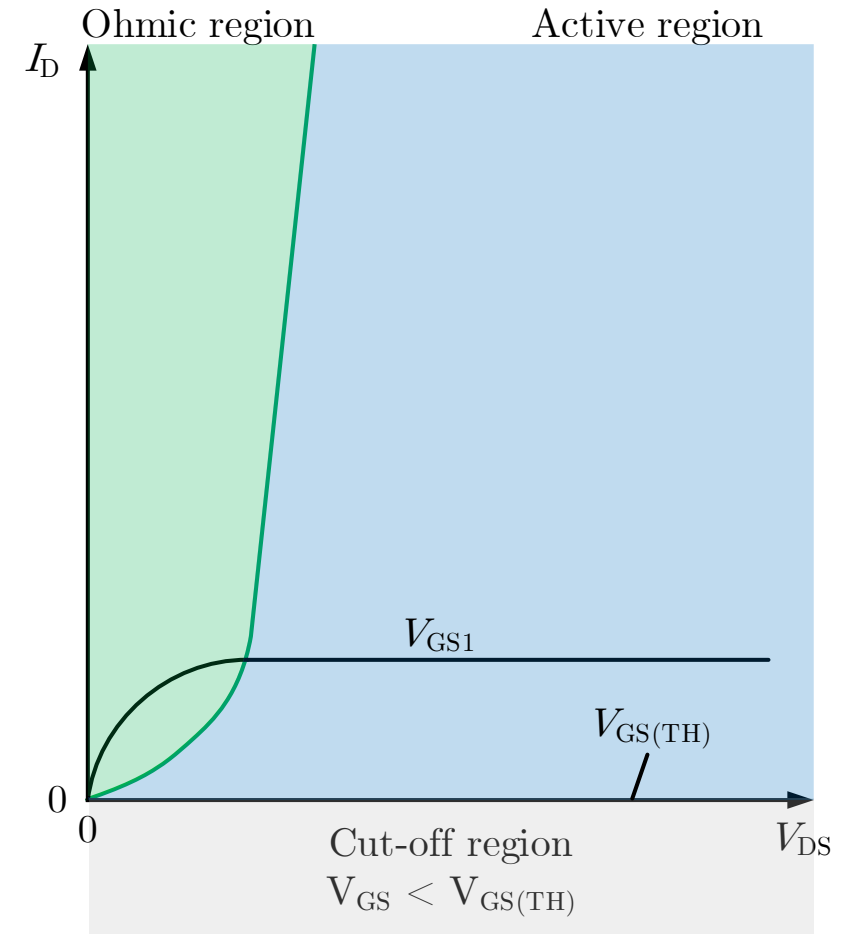
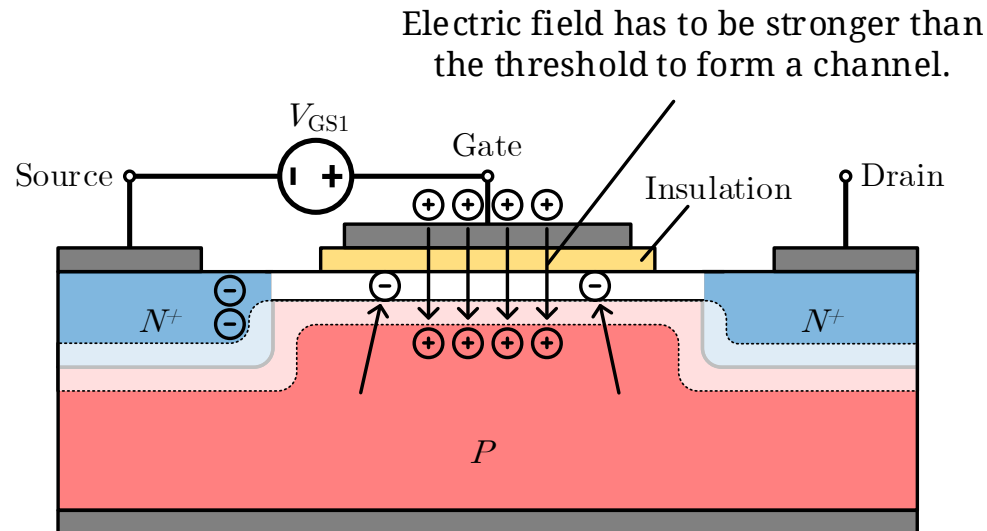
# MOSFETs

## Characteristics



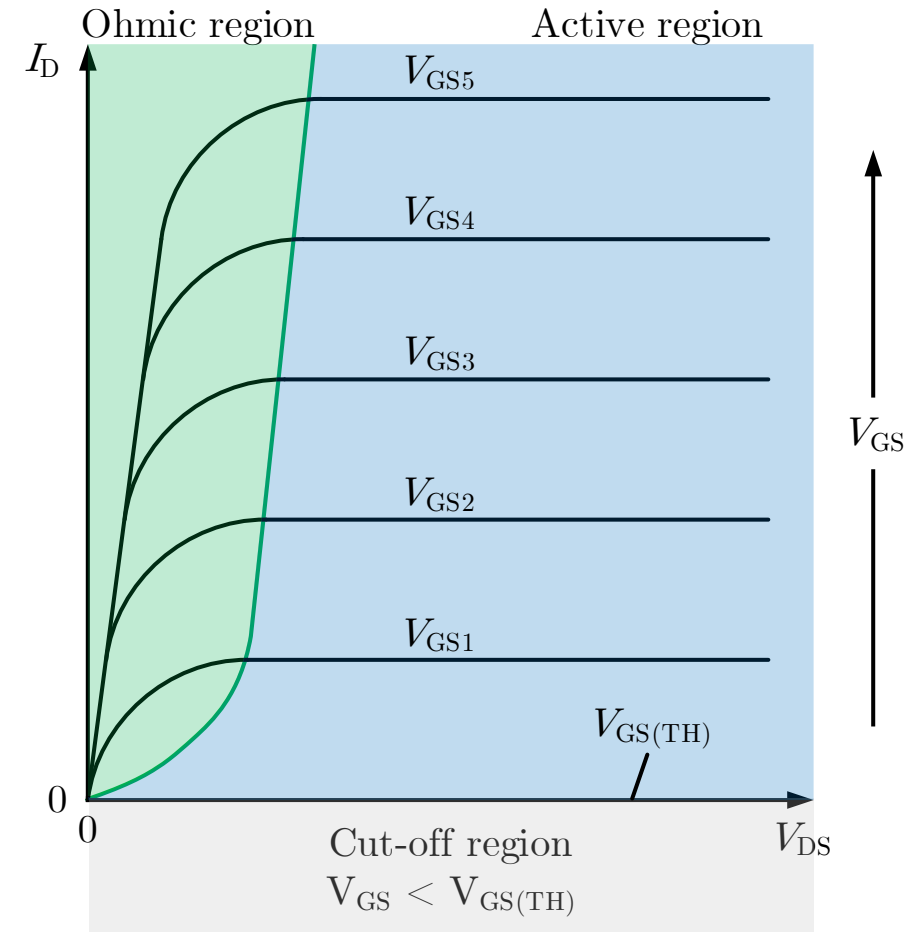
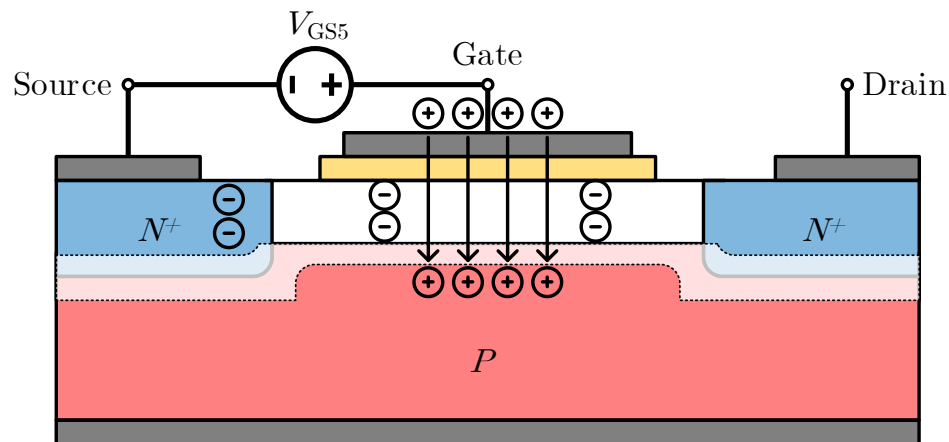
# MOSFETs – operating regions

- A sufficient amount of charge is needed at the gate to form a channel between the source and drain.
- Gate threshold voltage ( $V_{GS(TH)}$ ) is the minimum voltage required to turn the MOSFET on.
- Cut-off region when  $V_{GS} \leq V_{GS(TH)}$ .
- Ohmic region when  $V_{DS}$  increases proportionally to  $I_D$ .
- Active region when  $I_D$  and  $V_{DS}$  are independent of each other.



# MOSFETs – operating regions

- As a switch, the MOSFET should pass as much current as possible when turned on.
- Larger  $V_{GS}$  results in more charge accumulating at the gate to generate a stronger electric field.
- The stronger the electric field, the larger the channel.
- Large channels allow for higher current to flow within the switch ( $I_D$ ).
- The control of current flow using the electric field is called “field-effect”.



# MOSFET – FQP13N10

- The MOSFET used in this project is **FQP13N10**.
- The datasheet can be referred to find the  $V_{GS(TH)}$  and different operating regions of the MOSFET.
- In this MOSFET, the maximum current ( $I_D$ ) of 12.8A flows when  $V_{GS}$  is at least 15V.
- For a short amount of time, a much higher  $I_D$  can be pulsed.
- Note the operating regions are drawn in log scale.

## Absolute Maximum Ratings $T_C = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	FQP13N10	Unit
$I_D$	Drain Current - Continuous ( $T_C = 25^\circ\text{C}$ )	12.8	A
	- Continuous ( $T_C = 100^\circ\text{C}$ )	9.05	A
$I_{DM}$	Drain Current - Pulsed (Note 1)	51.2	A
$V_{GS}$	Gate-Source Voltage	$\pm 25$	V

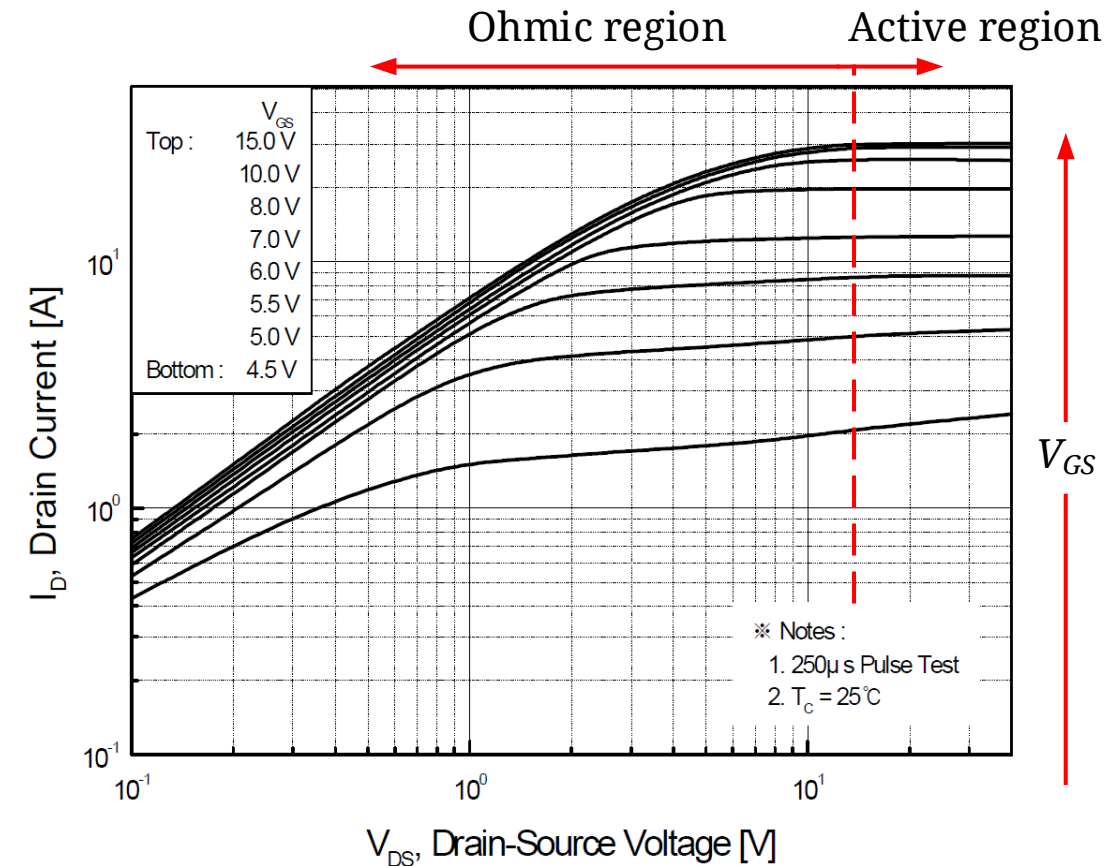
## Electrical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
<b>On Characteristics</b>						
$V_{GS(th)}$	Gate Threshold Voltage	$V_{DS} = V_{GS}, I_D = 250 \mu\text{A}$	2.0	--	4.0	V

### Notes:

1. Repetitive Rating : Pulse width limited by maximum junction temperature.

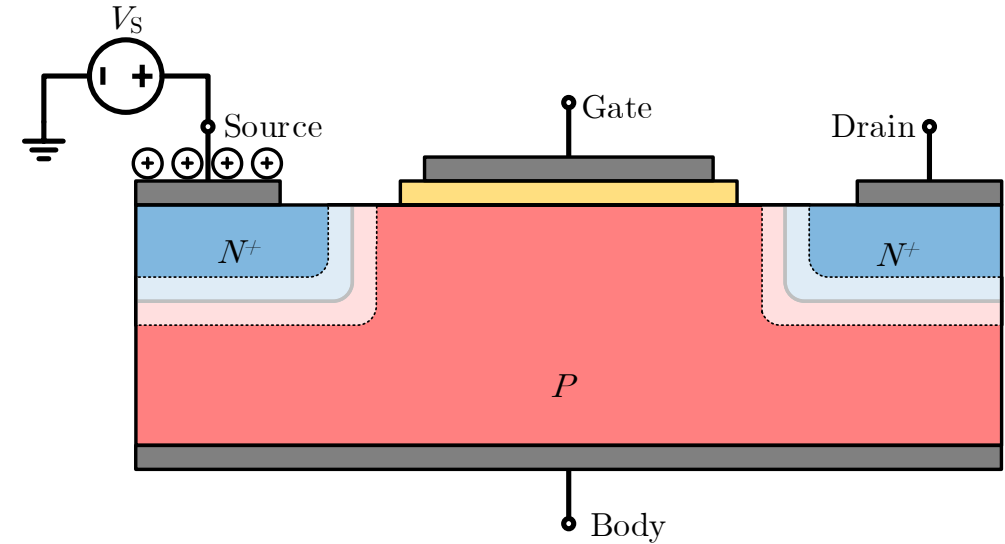
- $I_D$  and  $V_{GS}$  characteristics from FQP13N10 datasheet [13]



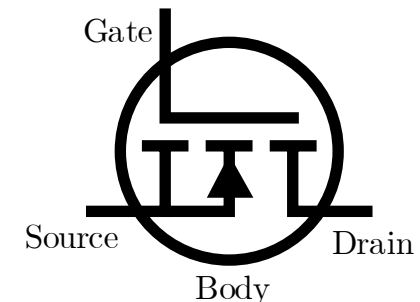
- On-region characteristics from FQP13N10 datasheet [13]

# MOSFETs – body and source

- The source and body are tied together for MOSFETs used for switching large currents.
- If voltage potential in the source is higher than the body, the depletion region increases and  $V_{GS(TH)}$  increases.
- If the MOSFET operates at a fixed  $V_{GS}$ , the maximum current through the drain ( $I_D$ ) is reduced due to a higher  $V_{GS(TH)}$ .
- In order to ensure  $I_D$  stays at the maximum level, the source and the body are tied together.
- Often the symbols will indicate that the source and body are tied together so there are only three terminals in the MOSFET instead of four.

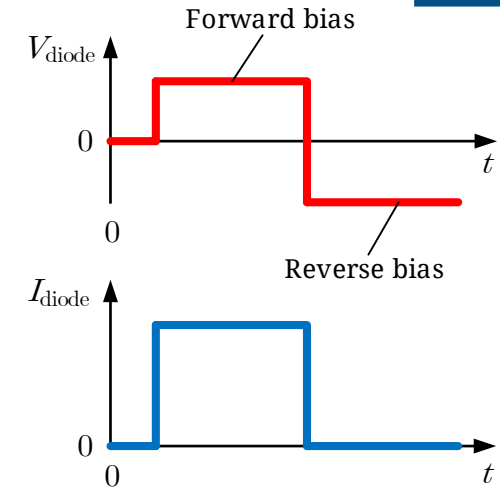
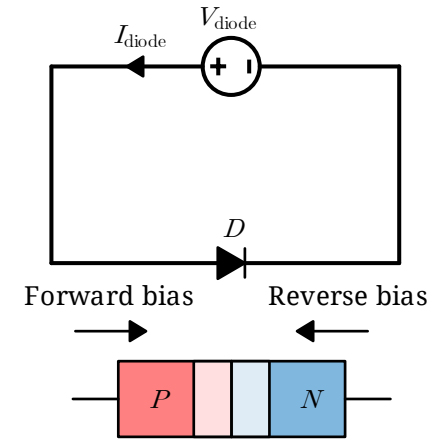


N-type enhancement

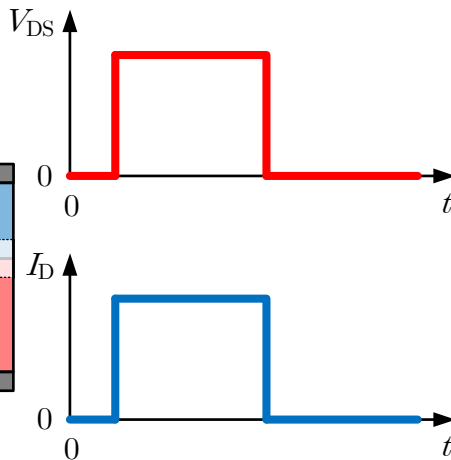
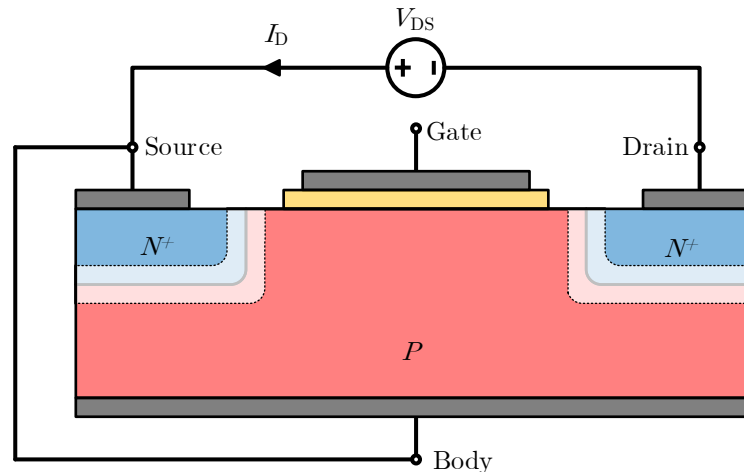
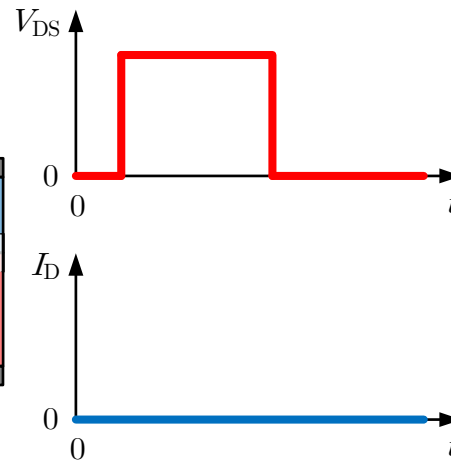
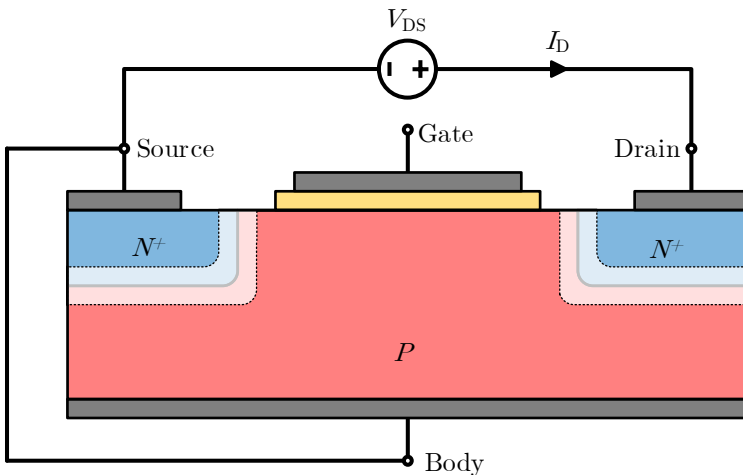


# MOSFETs – body diode

- When the MOSFET is forward biased, no current flows without positive voltage on the gate.
- When the MOSFET is reverse biased, current flows without any voltage to the gate.
- This one directional current flow acts like a diode so it is called a 'body diode'.

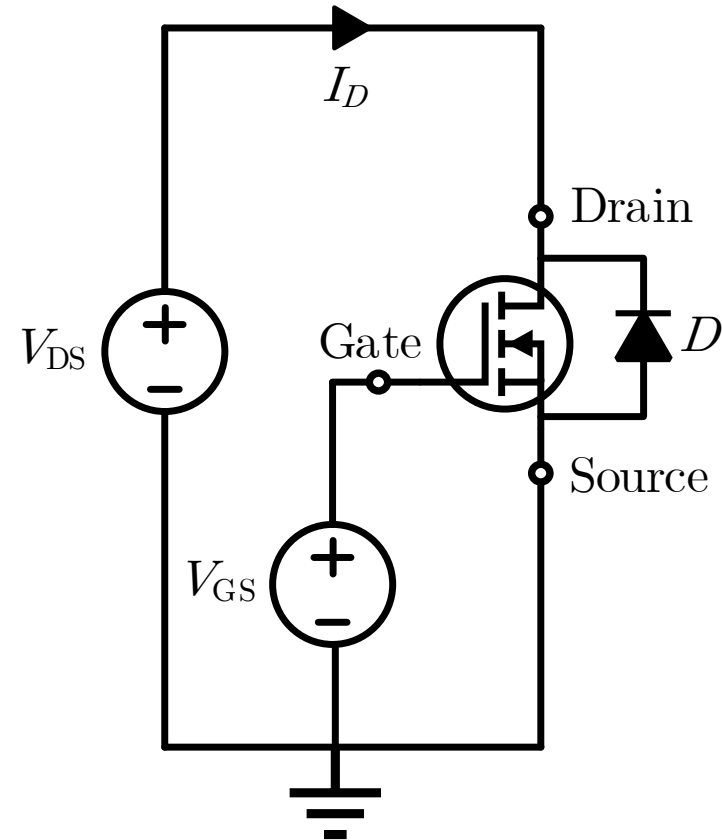


• Operation of a diode



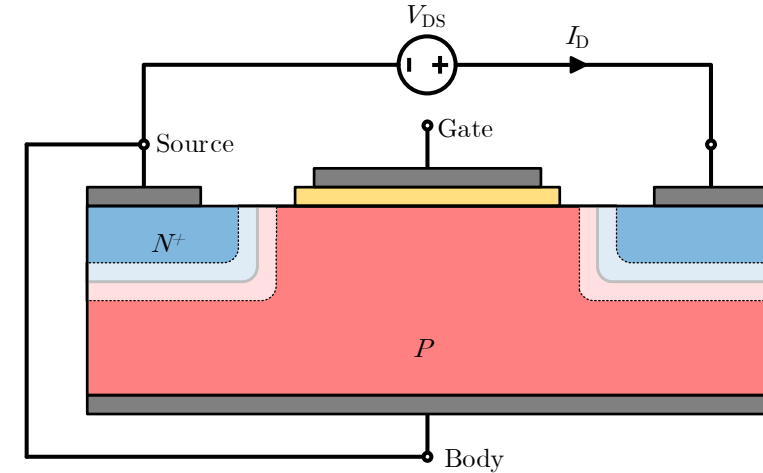
# MOSFETs – body diode

- MOSFET body diode is often drawn in the diagrams to include the current flow in the reverse direction.



# MOSFETs – breakdown voltages

- The maximum voltage that can be applied to the switch across the drain and source ( $V_{DSS}$ ) is listed in the absolute maximum ratings as 100V for FQP13N10.
- If  $V_{DSS(MAX)} > 100V$ , the MOSFET shorts across the drain and source so current flows freely and the MOSFET does not work like a switch.
- The **drain source breakdown voltage** ( $BV_{DSS}$ ) is also listed in the off-characteristics to show the maximum rating of the switch again.
- Usually, switches are chosen to operate well within the absolute maximum ratings. In actual circuits, transients in the system cause voltage spikes that may exceed the calculated voltages for the switches.
- As a rule of thumb, select switches that has a breakdown voltage at least 30% higher than the maximum voltage stress.



## Absolute Maximum Ratings $T_C = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	FQP13N10	Unit
$V_{DSS}$	Drain-Source Voltage	100	V

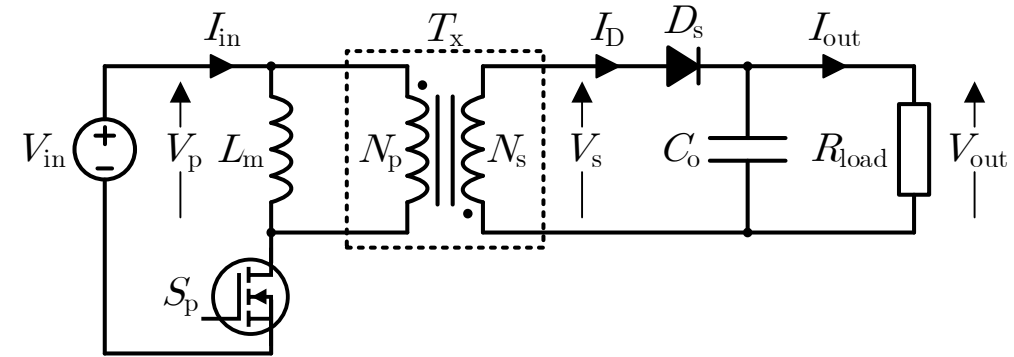
## Electrical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
$BV_{DSS}$	Drain-Source Breakdown Voltage	$V_{GS} = 0\text{ V}, I_D = 250\text{ }\mu\text{A}$	100	--	--	V

- Breakdown voltages from FQP13N10 datasheet [13]

# MOSFETs – example flyback converter

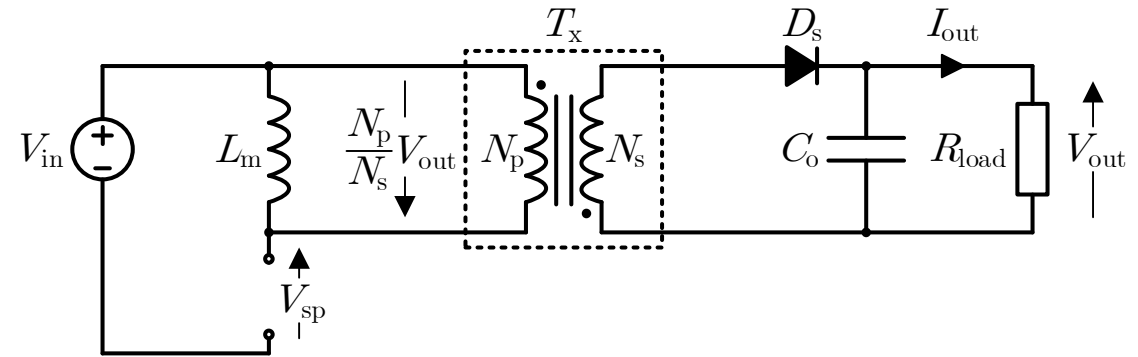
- Example flyback converter parameters:
- Input voltage ( $V_{in}$ ): 30V
- Output voltage ( $V_{out}$ ): 30V
- Turns ratio ( $\frac{N_p}{N_s}$ ): 1
- Operating frequency: 100kHz
- Duty cycle: 0.5
- MOSFET maximum voltage: 100V
- Coupling factor = 1
- Discontinuous conduction mode





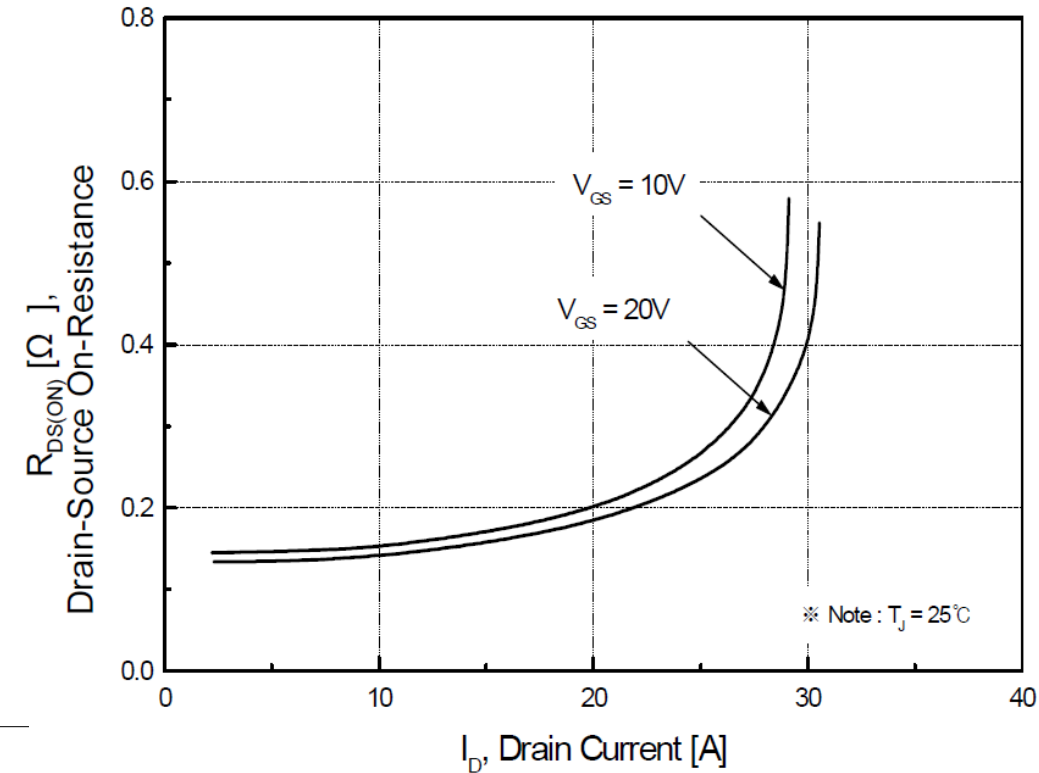
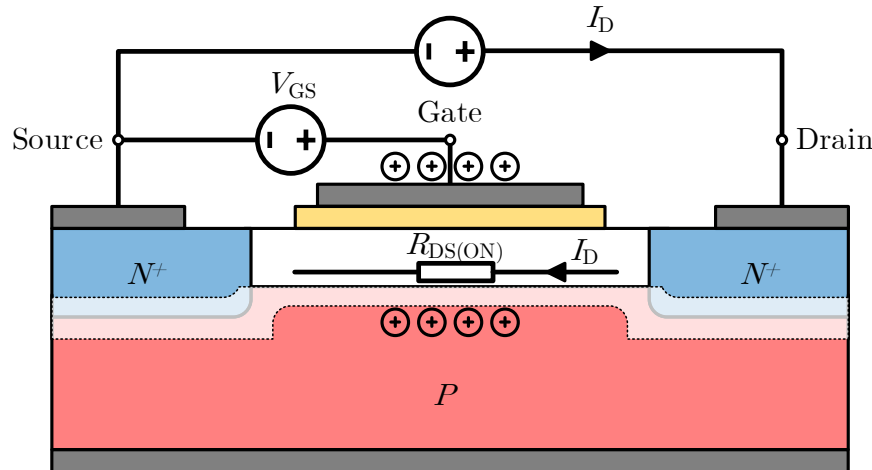
# MOSFETs – switch voltage

- When the switch is off, the switch sees the input voltage ( $V_{in}$ ) and output voltage ( $V_{out}$ ).
- Output voltage on the secondary is kept up by the capacitor. Secondary voltage is reflected to the primary by the turns ratio ( $\frac{N_p}{N_s}$ ).
- Switch voltage is given by:
- $V_{sp(max)} = V_{in} + \frac{N_p}{N_s} V_{out}$
- $\therefore V_{sp(max)} = 30 + 1 \times 30 = 60V$
- FQP13N10 has absolute maximum rating of 100V across the switch.
- Check the simulations for any transients above 100V.



# MOSFETs – on-characteristics

- On-resistance ( $R_{DS(ON)}$ ) includes all resistances in the MOSFET in the path of current flow.
- A higher  $V_{GS}$  provides a wider channel to reduce  $R_{DS(ON)}$ .
- A trade-off between shorter channel length reducing  $R_{DS(ON)}$  and longer channel lengths improving breakdown voltage.



$R_{DS(on)}$	Static Drain-Source On-Resistance	$V_{GS} = 10\text{ V}, I_D = 6.4\text{ A}$	--	0.142	0.18	$\Omega$
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- $I_D$  and  $V_{GS}$  characteristics from FQP13N10 datasheet [13]

- On-region characteristics from FQP13N10 datasheet [13]

# MOSFETs – conduction loss

- Switch RMS current is given by:

$$I_{sw,rms} = \sqrt{\frac{1}{t_s} \int_0^{t_s} I^2(t) dt}$$

- Here, when switch is conducting:

$$I(t) = \frac{I_{sw,peak}}{Dt_s} t$$

$$I_{sw,rms} = \sqrt{\frac{1}{t_s} \int_0^{Dt_s} \frac{I_{sw,peak}^2 t^2}{D^2 t_s^2} dt} = \frac{I_{sw,peak}^2 D^3 t_s^3}{3D^2 t_s^3} = \sqrt{\frac{D}{3}} I_{sw,peak}$$

$$I_{sw,rms} = \sqrt{\frac{D}{3}} I_{sw,peak}$$

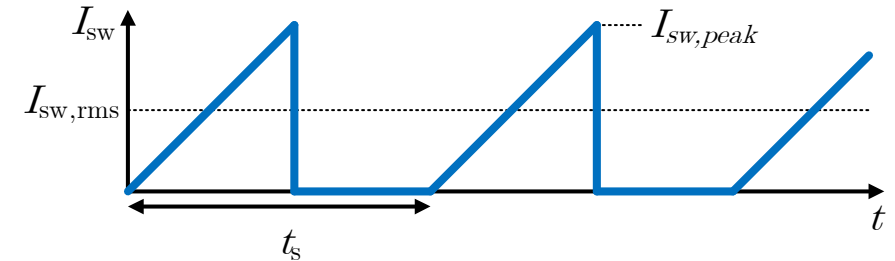
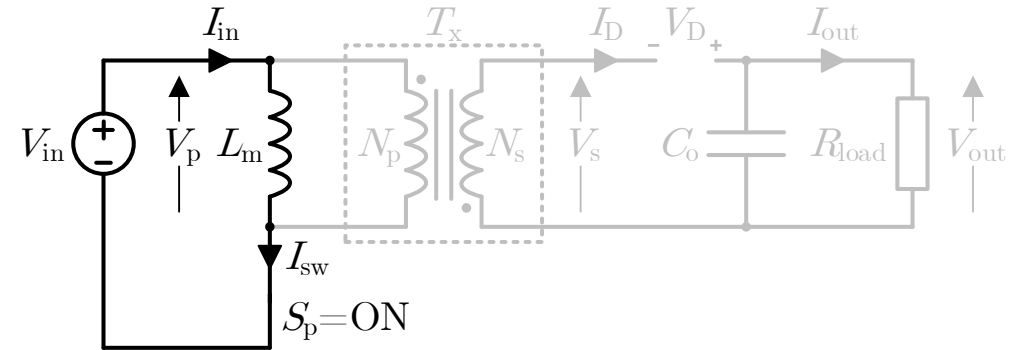
- In this example,  $I_{sw,peak} = 4A$  and  $R_{ds,on} = 0.142m\Omega$ .

$$I_{sw,rms} = \sqrt{\frac{0.5}{3}} \times 4 = 1.63A$$

- Conduction loss is given by:

$$P_{cond} = I_{sw,rms}^2 R_{ds,on}$$

$$P_{cond} = 1.63^2 \times 0.142 = 377mW$$

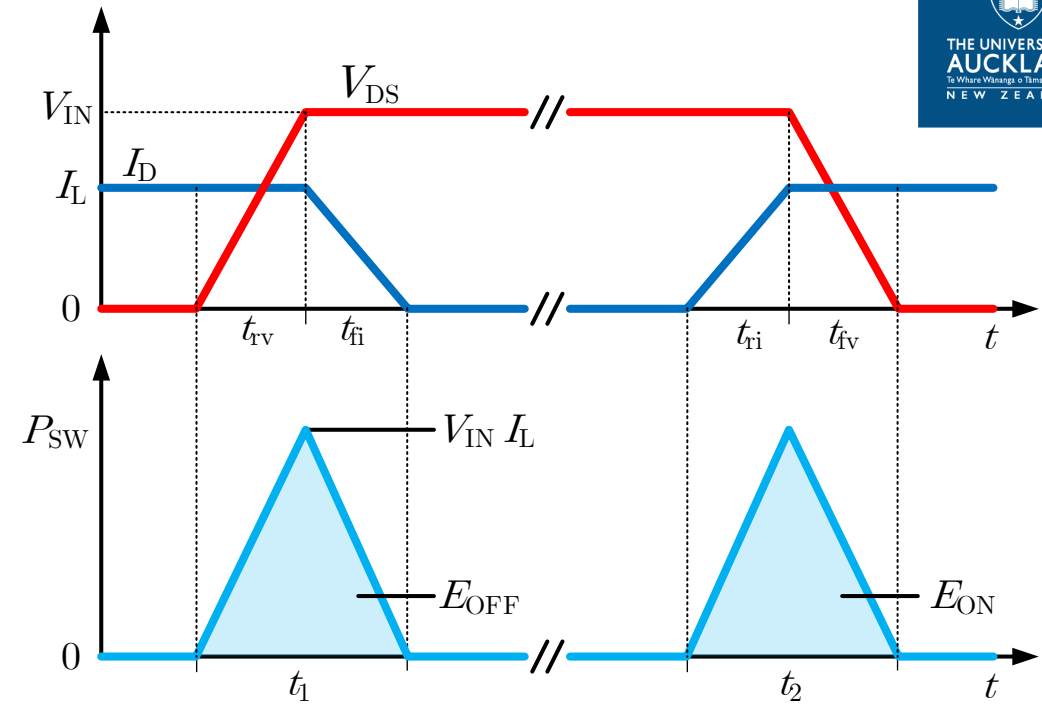


$R_{DS(on)}$	Static Drain-Source On-Resistance	$V_{GS} = 10\text{ V}, I_D = 6.4\text{ A}$	--	0.142	0.18	$\Omega$
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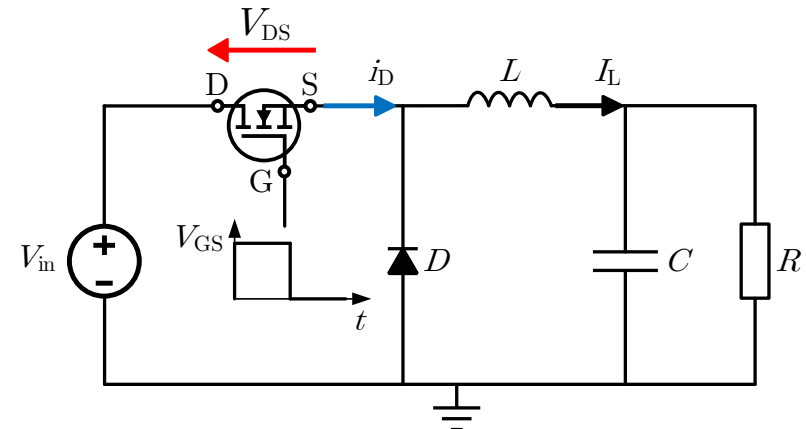
- On-resistance from FQP13N10 datasheet [13]

# MOSFETs – switching loss

- Switching losses ( $P_{SW}$ ) is the power lost when the MOSFET turns on or off.
- A very simplified example is shown here.
- Energy lost ( $E$ ) =  $\frac{1}{2} V_{in} I_L (t_{rv} + t_{fi}) + \frac{1}{2} V_{in} I_L (t_{ri} + t_{fv})$
- Energy lost ( $E$ ) =  $E_{ON} + E_{OFF}$
- Switching loss ( $P_{SW}$ ) =  $\frac{1}{T} E_{TOTAL} = (E_{ON} + E_{OFF})f$
- Assume  $t_{rv} = t_{fi}$  and  $t_{ri} = t_{fv}$
- $\therefore P_{SW} = (V_{in} I_L t_{fi} + V_{in} I_L t_{ri})f$



- Example switching waveforms when MOSFET turns on or off



- An example circuit using a MOSFET

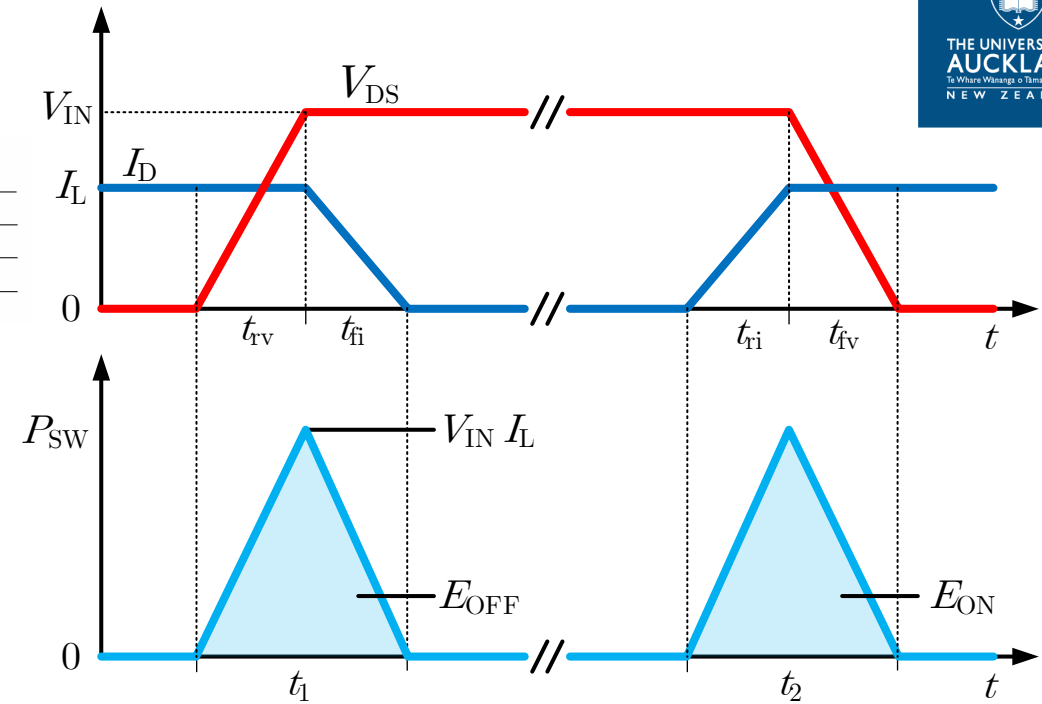
# MOSFETs – switching loss

## Switching Characteristics

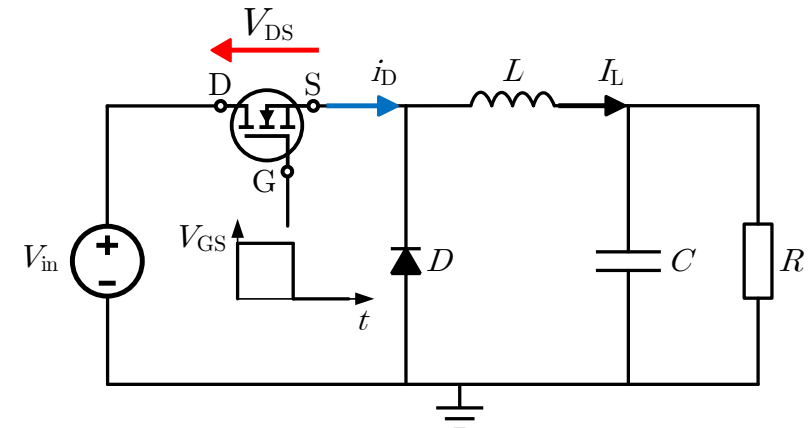
$t_{d(on)}$	Turn-On Delay Time	$V_{DD} = 50\text{ V}, I_D = 12.8\text{ A},$ $R_G = 25\ \Omega$	--	5	20	ns
$t_r$	Turn-On Rise Time		--	55	120	ns
$t_{d(off)}$	Turn-Off Delay Time		--	20	50	ns
$t_f$	Turn-Off Fall Time		--	25	60	ns

(Note 4)

- Switching characteristics from FQP13N10 datasheet [13]
- In this example,  $V_{in} = 30\text{V}$ ,  $I_{sw,peak} = 4\text{A}$  and  $f = 100\text{kHz}$ .
- $t_{ri} = t_{fv} = 55\text{ns}$
- $t_{fi} = t_{rv} = 25\text{ns}$
- $P_{SW} = (V_{in}I_L t_{fi} + V_{in}I_L t_{ri})f$
- $P_{SW} = (30 \times 4 \times 55\text{n} + 30 \times 4 \times 25\text{n})100\text{k}$
- $P_{SW} = 960\text{mW}$
- Switching losses tend to be much higher than conduction losses.
- Switching losses are proportional to switching frequency.
- Note that in the flyback converter operating in discontinuous conduction mode, the current is 0A at turn on.



- Example switching waveforms when MOSFET turns on or off



- An example circuit using a MOSFET

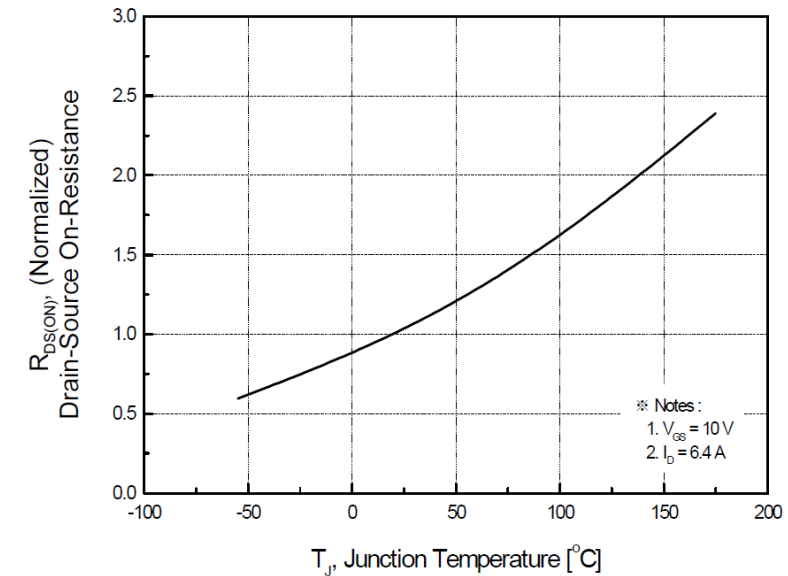
# MOSFETs – thermal characteristics

- As the temperature of the MOSFET increases,  $R_{DS(ON)}$  also increases that leads to higher power loss.
- The capacity of the MOSFET to handle  $I_D$  also reduces as the temperature increases.
- Care should be taken to not exceed the maximum junction temperature of 175C as listed in the datasheet for FQP13N10.

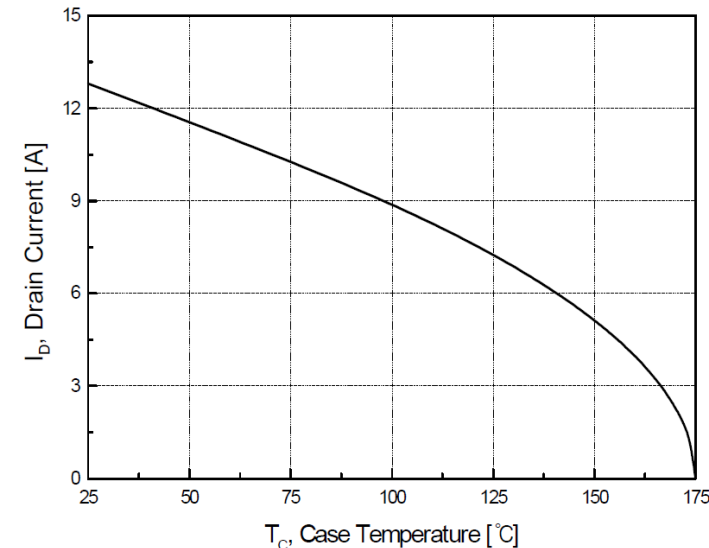
## Thermal Characteristics

Symbol	Parameter	FQP13N10	Unit
$R_{\theta JC}$	Thermal Resistance, Junction-to-Case, Max.	2.31	°C/W
$R_{\theta JA}$	Thermal Resistance, Junction-to-Ambient, Max.	62.5	°C/W

- Thermal resistance from FQP13N10 datasheet [13]
- Given the information in the datasheet, the maximum power dissipation in the MOSFET can be estimated.
- For ambient temperature of 25C:
- $\frac{1}{R_{\theta JA}} (T_{\text{junction}} - T_{\text{ambient}}) = P_{\text{MAX}}$
- $P_{\text{MAX}} = \frac{1}{62.5} (175 - 25) = 2.4W$



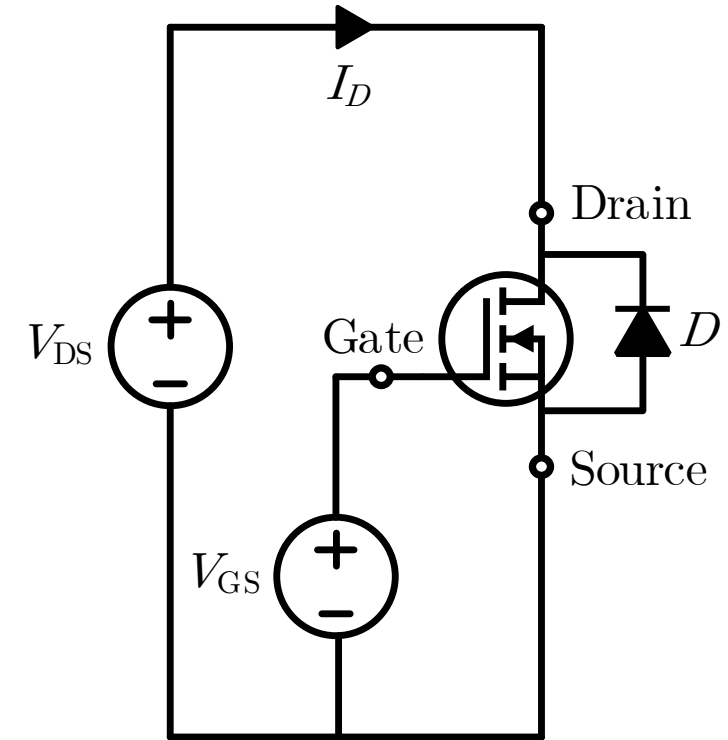
- On-resistance variation against temperature [13]



- Maximum drain current against temperature

# MOSFETs – summary

- Different operating regions – Ohmic, active and cut-off.
- Source to body connection and body diode.
- Breakdown voltages: 100V for FQP13N10.
- Conduction losses:  $P_{\text{cond}} = I_{\text{sw,rms}}^2 R_{\text{ds,on}}$ .
- $R_{\text{ds,on}}$  decreased when  $V_{\text{GS}}$  is higher.
- Usually a trade-off between breakdown voltages and  $R_{\text{ds,on}}$ .
- Switching losses:  $P_{\text{SW}} = (E_{\text{ON}} + E_{\text{OFF}})f = P_{\text{SW}} = (V_{\text{in}}I_{\text{L}}t_{\text{fi}} + V_{\text{in}}I_{\text{L}}t_{\text{ri}})f$
- Maximum junction temperature: 175°C for FQP13N10.



- An example N-type enhancement MOSFET circuit

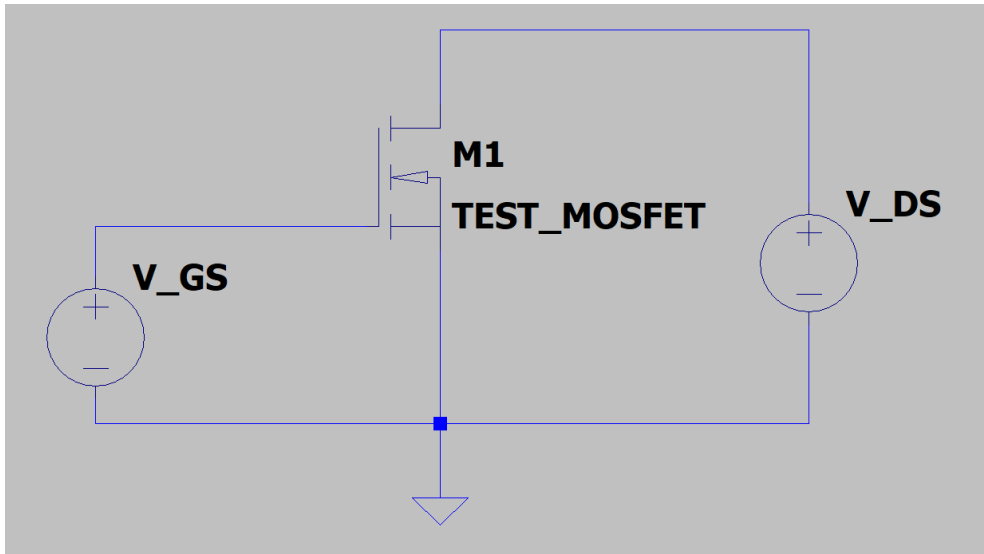
# MOSFETs

## Drivers and simulation examples

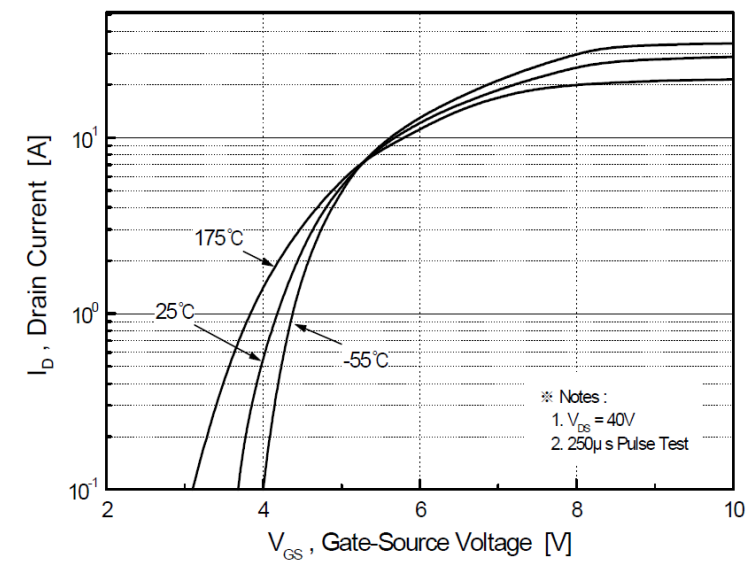


# MOSFETs – simulation model

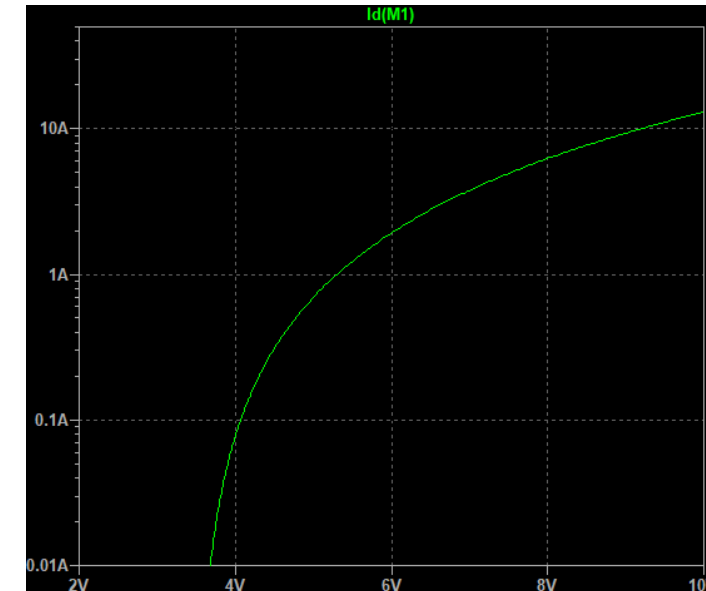
- A simple MOSFET model can be made on LTSpice to approximate the operation.
- More accurate simulation models quickly get much more complicated and involves a lot of empirical data.
- For the purposes of this course, the simulation model can be downloaded from [14].



- Simulation of a simple MOSFET model

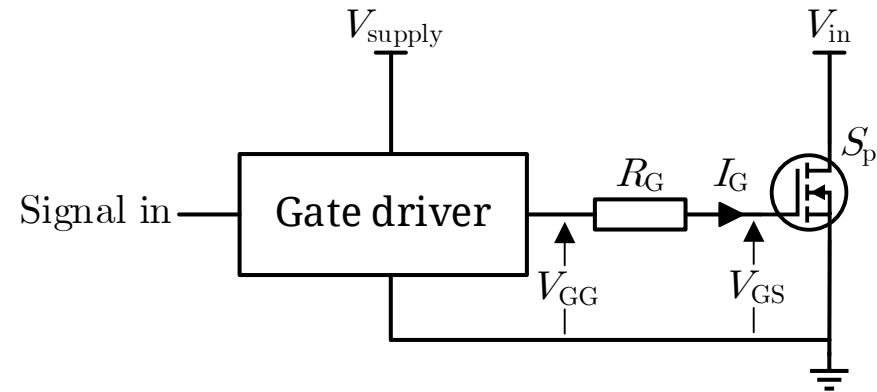
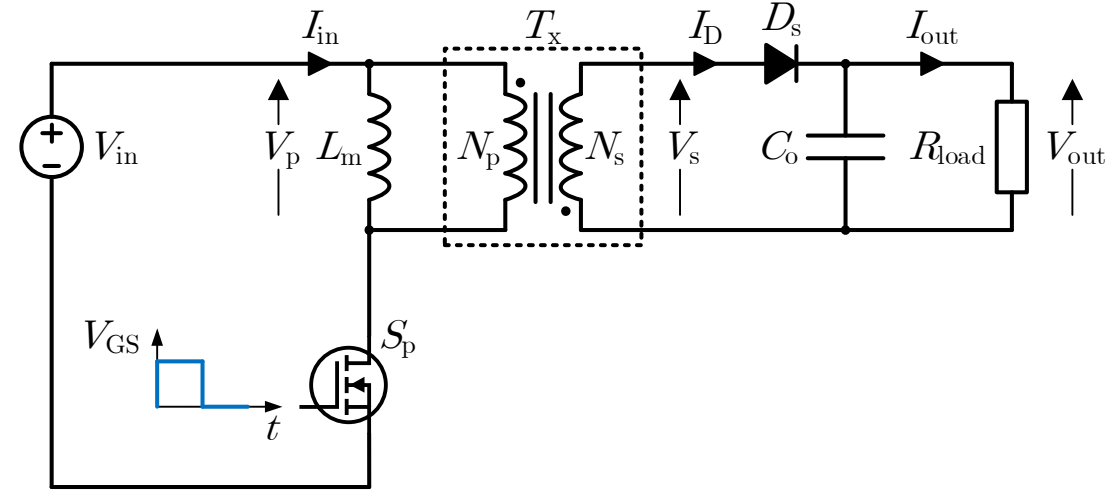


- Transfer characteristics [13]



# MOSFETs – drivers

- Generally, a separate circuit called **MOSFET drivers** are used to turn MOSFETs on and off.
- MOSFET drivers are specifically designed to achieve high voltage across the gate and source ( $V_{GS}$ ) and current into the gate ( $I_G$ ) that may not be achievable by using a standard controller circuit.
- MOSFET driver is needed to:
  - Achieve high  $V_{GS}$  to increase  $I_D$  flow through the switch as shown before in the characteristics curve.
  - High  $V_{GS}$  helps to improve the conduction loss as  $R_{DS(ON)}$  is reduced.
  - Achieve high current into the gate ( $I_G$ ) helps to charge the gate faster for to enable higher switching frequency.
  - Faster switching times helps to reduce switching losses.
- MOSFET driver can be considered as a ‘power amplifier’ for the gate.



# MOSFETs – UC3843 driver

- The MOSFET is driven by UC3843 chip in this project.
- Ideally, there would be no gate resistance from the MOSFET driver to the gate of the MOSFET.
- Sometimes, the switching is too fast and needs to be slowed down to reduce transients (overshoots and ringing) caused by parasitics that could damage the circuit.
- Gate resistance can be approximated by:
- $R_G = \frac{V_{GG}}{I_G}$
- If  $V_{GG} = 15V$  and  $I_G = 1A$ ,
- $R_G = \frac{15}{1} = 15\Omega$
- Here, 1A is from the UC3843 datasheet.
- In the practical circuit, the value may require some adjustment as parasitics are difficult to predict.

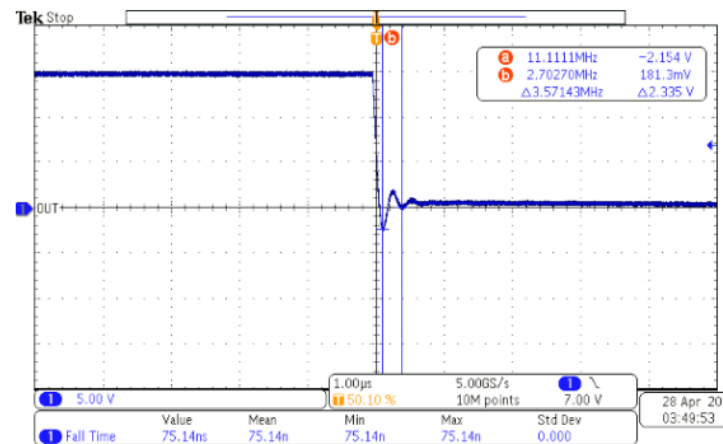
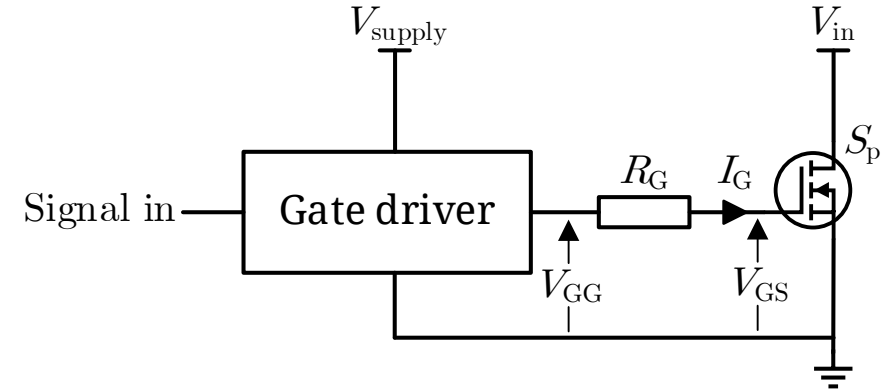


Figure 4. External Gate Resistor  $R_{GATE} = 0\text{-}\Omega$

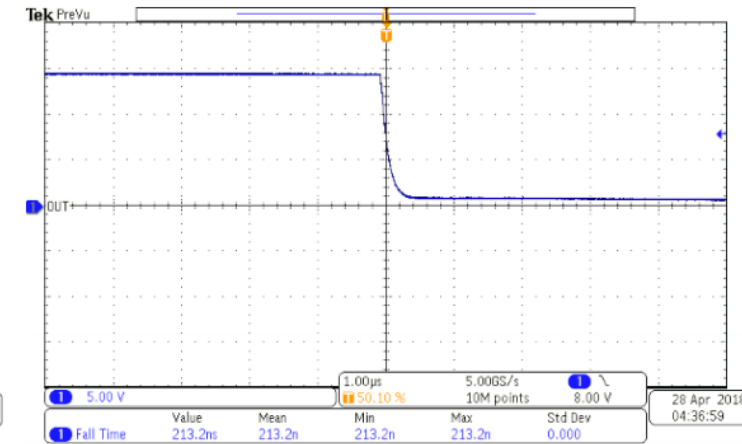


Figure 5. Critically Damped External Gate Resistor  $R_{GATE} = 7\text{-}\Omega$

- Scope measurements of a MOSFET with and without gate resistance.



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# Appendix

# MOSFETs – safe operating area

- Maximum voltage and current defined in the datasheet as 100V and 51.2A (pulsed).

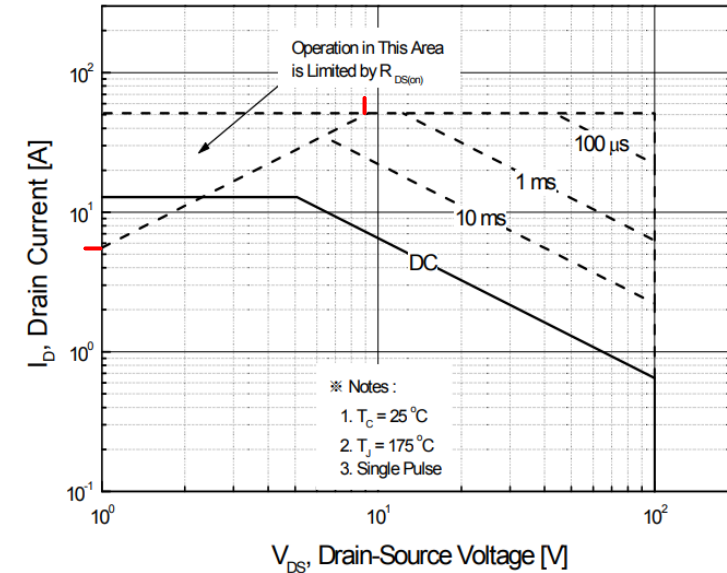
## Absolute Maximum Ratings T<sub>C</sub> = 25°C unless otherwise noted.

Symbol	Parameter	FQP13N10	Unit
V <sub>DSS</sub>	Drain-Source Voltage	100	V
I <sub>D</sub>	Drain Current	- Continuous (T <sub>C</sub> = 25°C)	12.8
		- Continuous (T <sub>C</sub> = 100°C)	9.05
I <sub>DM</sub>	Drain Current - Pulsed	(Note 1)	51.2

## Electrical Characteristics T<sub>C</sub> = 25°C unless otherwise noted.

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
<b>On Characteristics</b>						
R <sub>DS(on)</sub>	Static Drain-Source On-Resistance	V <sub>GS</sub> = 10 V, I <sub>D</sub> = 6.4 A	--	0.142	0.18	Ω

- Operation is limited by the maximum R<sub>DS(on)</sub>.
- At V<sub>DS</sub> = 1V,  $I_D = \frac{V_{DS}}{R_{DS(on),max}} = \frac{1}{0.18} = 5.56A$
- At I<sub>D</sub> = 51.2A, V<sub>DS</sub> = I<sub>D</sub>R<sub>DS(on),max</sub> = 51.2 × 0.18 = 9.22V
- Switch can only operate within the region bound by these two points.



- Maximum safe operating area [13]

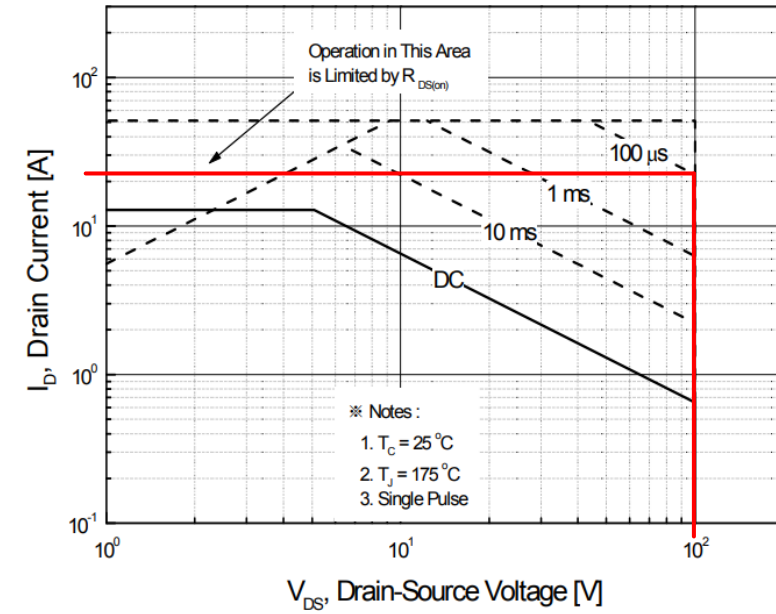
- On-resistance variation vs temperature [13]

# MOSFETs – safe operating area

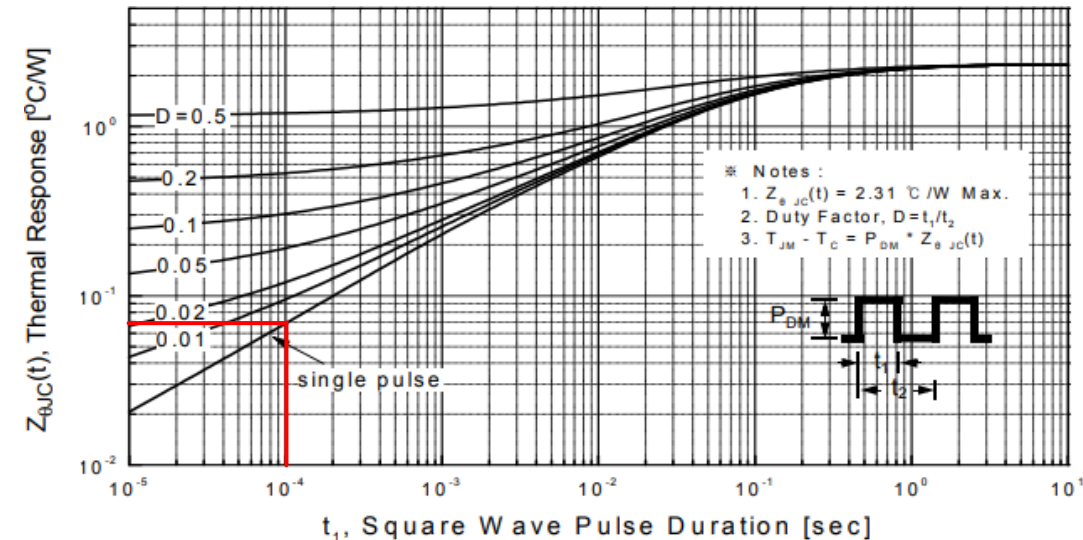
- According to the datasheet:  $T_{j(max)} - T_c = P_{D(max)} \times Z_{\theta jc}$
- Here,  $P_{D(max)}$  is the power loss during the pulse and  $Z_{\theta jc}$  is the thermal response in  $^{\circ}\text{C}/\text{W}$ .
- For example, a single pulse of  $100\mu\text{s}$  has  $Z_{\theta jc} \approx 0.07^{\circ}\text{C}/\text{W}$ .
- If  $V_{DS} = 100\text{V}$ , what is the maximum pulse current?
- $T_{j(max)} = 175^{\circ}\text{C}$  and  $T_c = 25^{\circ}\text{C}$
- $175 - 25 = 100 \times I_D \times 0.07$
- $\frac{150}{100 \times 0.07} = 21.4\text{A}$

## Absolute Maximum Ratings $T_c = 25^{\circ}\text{C}$ unless otherwise noted.

Symbol	Parameter	FQP13N10	Unit
$P_D$	Power Dissipation ( $T_c = 25^{\circ}\text{C}$ )	65	W
	- Derate above $25^{\circ}\text{C}$	0.43	$\text{W}/^{\circ}\text{C}$
$T_J, T_{STG}$	Operating and Storage Temperature Range	-55 to +175	$^{\circ}\text{C}$



- Maximum safe operating area [13]



- Transient thermal response curve [13]

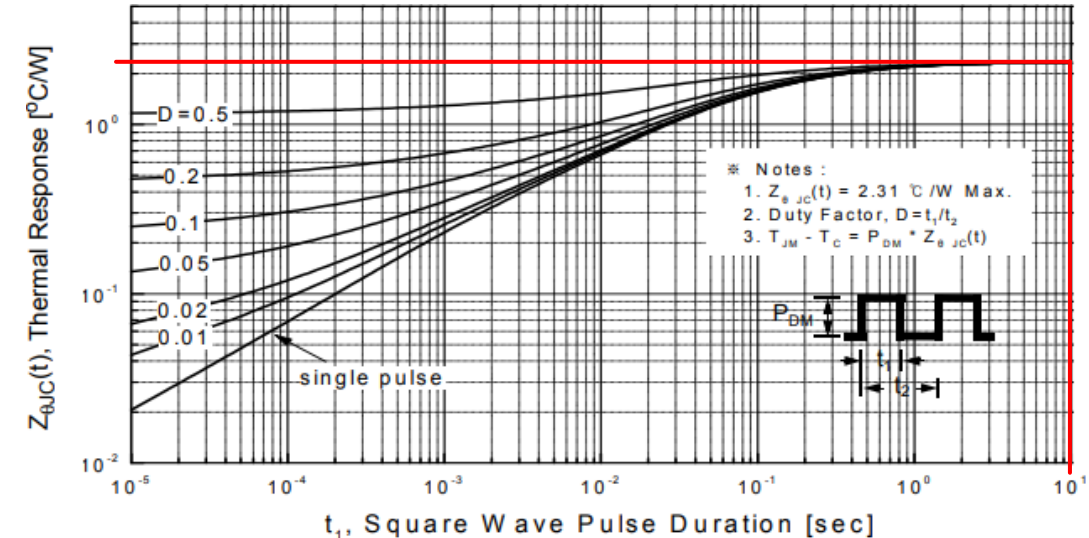
# MOSFETs – safe operating area

- When duty cycle is increased,  $Z_{\theta jc}$  is higher so  $P_D$  gets smaller.
- At the other end of the spectrum, if  $D = 0.5$  and pulse is 10s long:  $Z_{\theta jc(max)} = 2.31^\circ\text{C/W}$
- $175 - 25 = P_{D(max)} \times 2.31$
- $\therefore P_{D(max)} = 65.0\text{W}$

## Absolute Maximum Ratings $T_C = 25^\circ\text{C}$ unless otherwise noted.

Symbol	Parameter	FQP13N10	Unit
$P_D$	Power Dissipation ( $T_C = 25^\circ\text{C}$ )	65	W
	- Derate above $25^\circ\text{C}$	0.43	W/ $^\circ\text{C}$
$T_J, T_{STG}$	Operating and Storage Temperature Range	-55 to +175	$^\circ\text{C}$

- The datasheet numbers here assume  $T_c$  stays constant at  $25^\circ\text{C}$ .  $Z_{\theta jc}$  will change in real life scenarios as  $T_c$  changes.

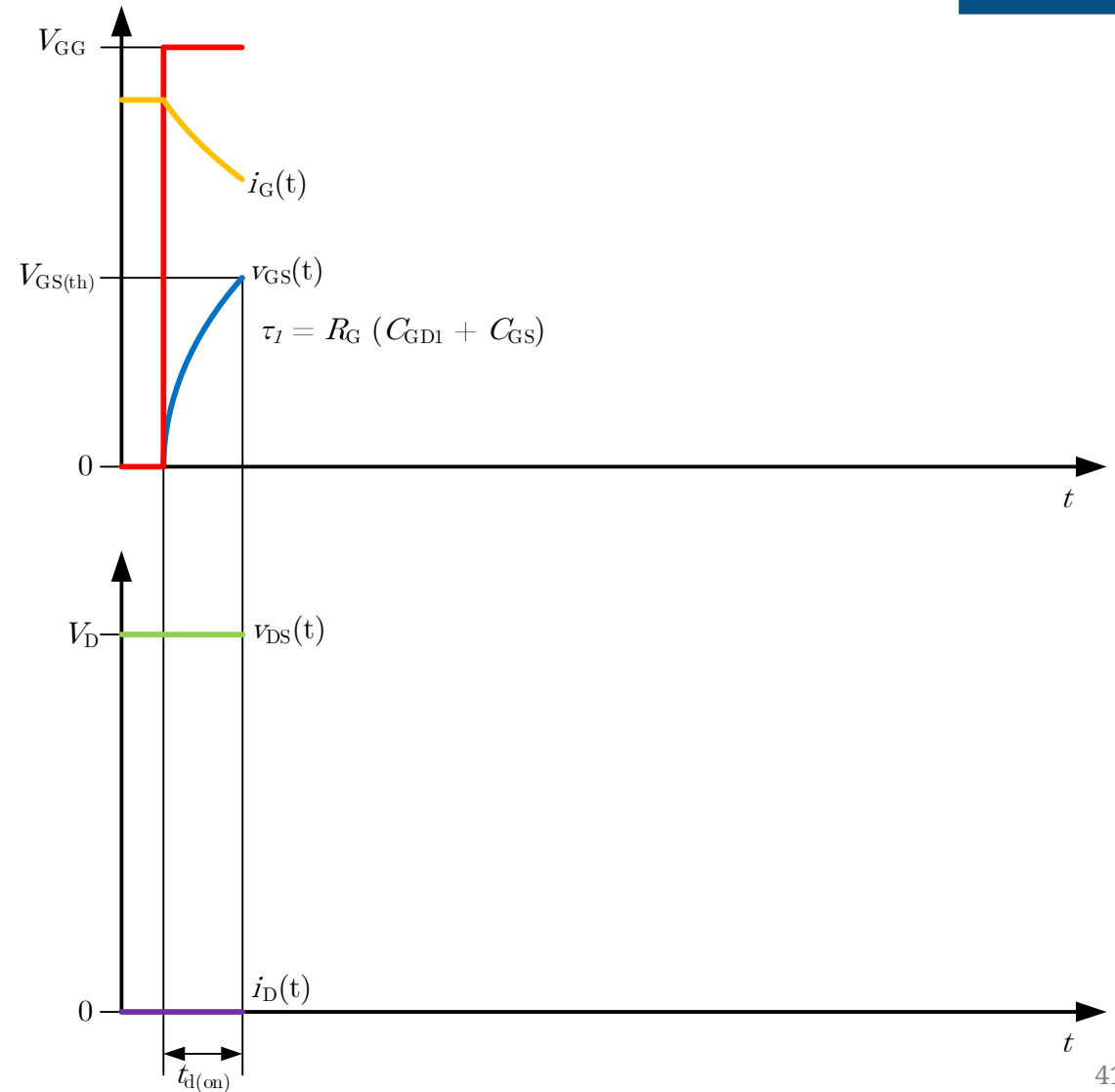
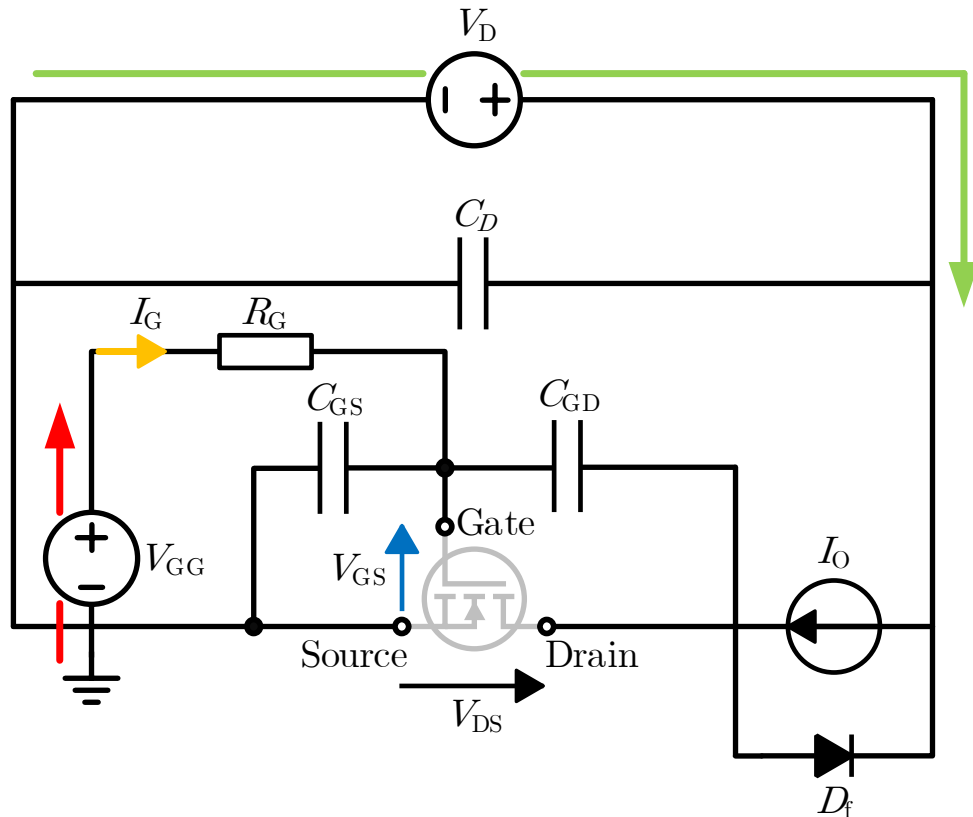


- Transient thermal response curve [13]



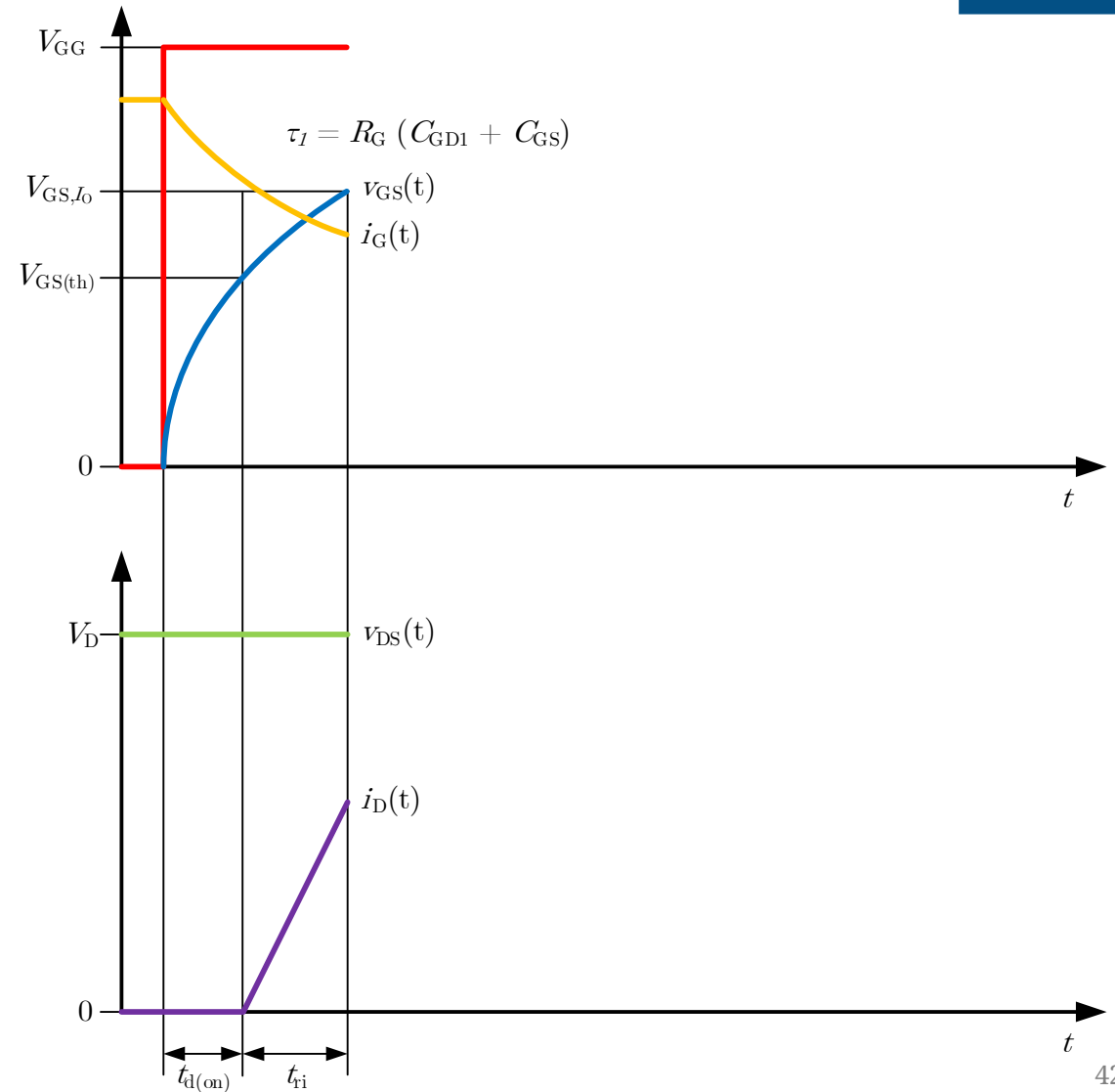
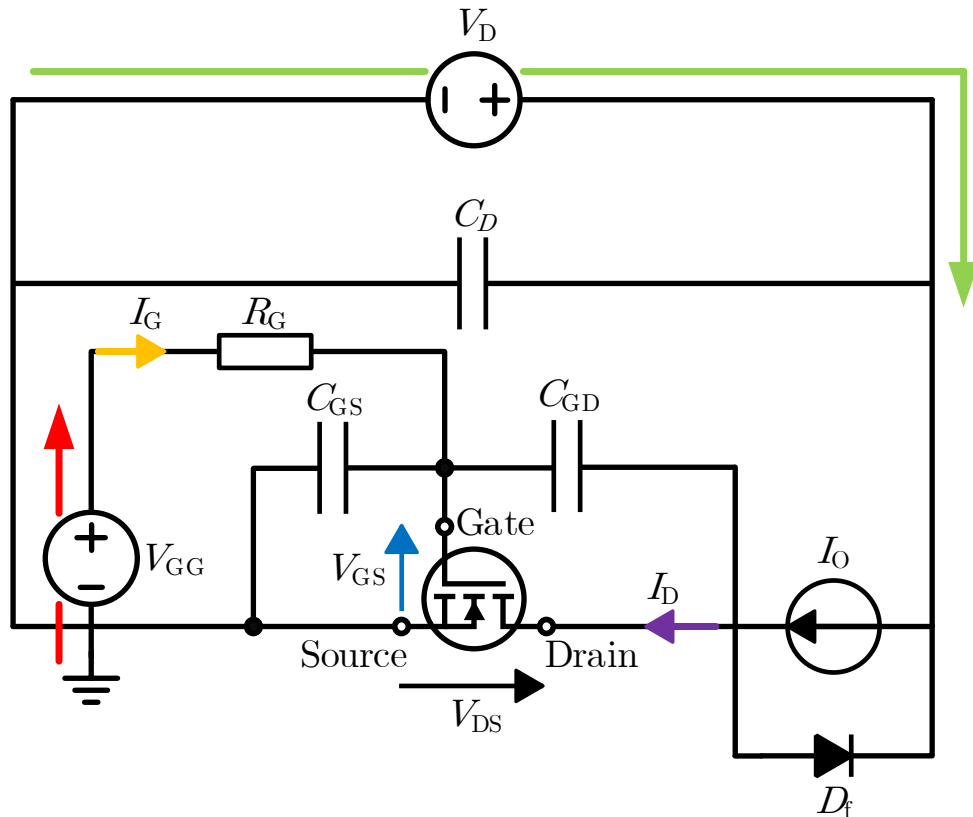
# MOSFETs – switching losses

- Voltage ( $V_{GG}$ ) applied at the gate.
- Capacitors ( $C_{GS}$  and  $C_{GD}$ ) charge up.



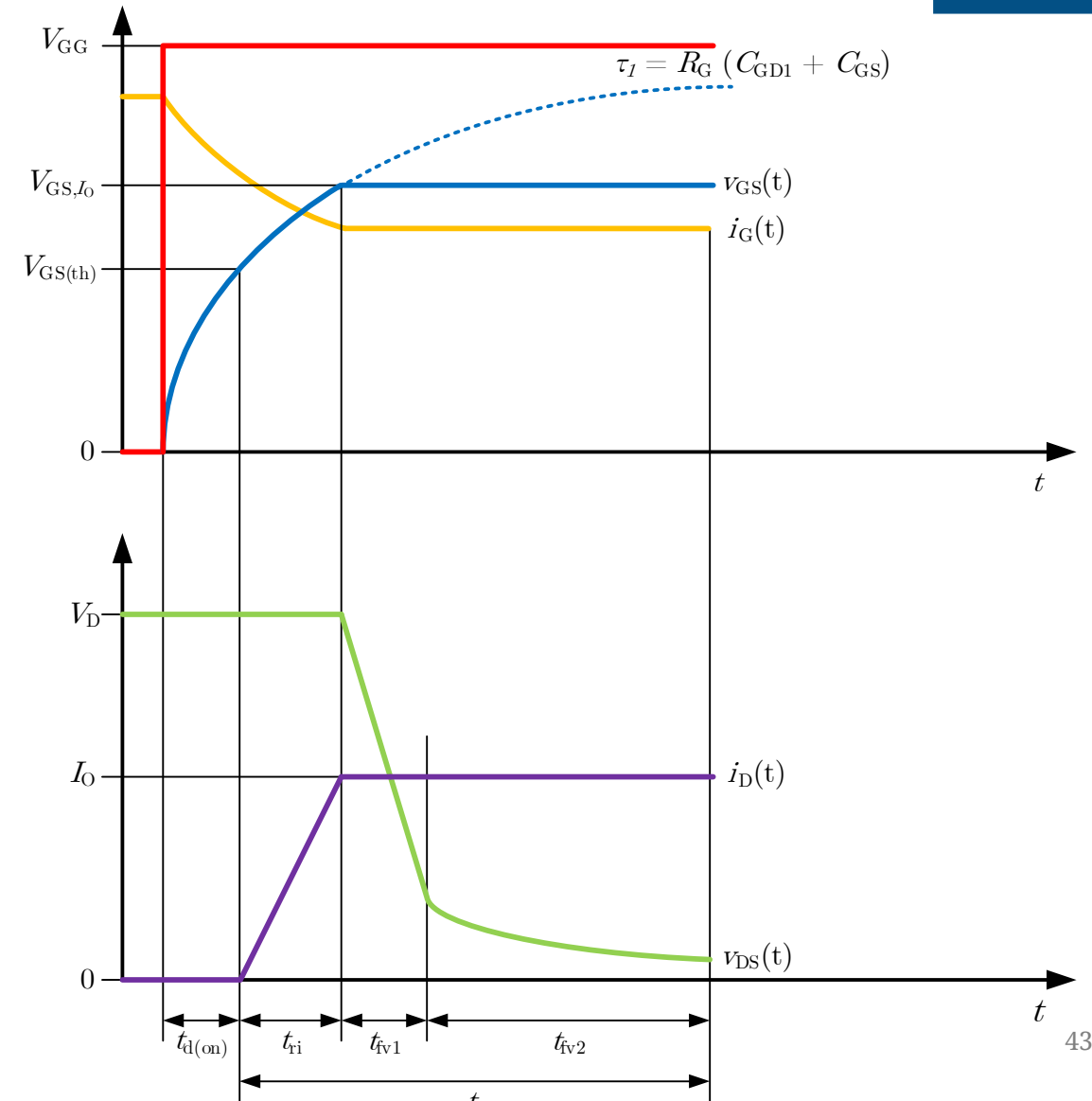
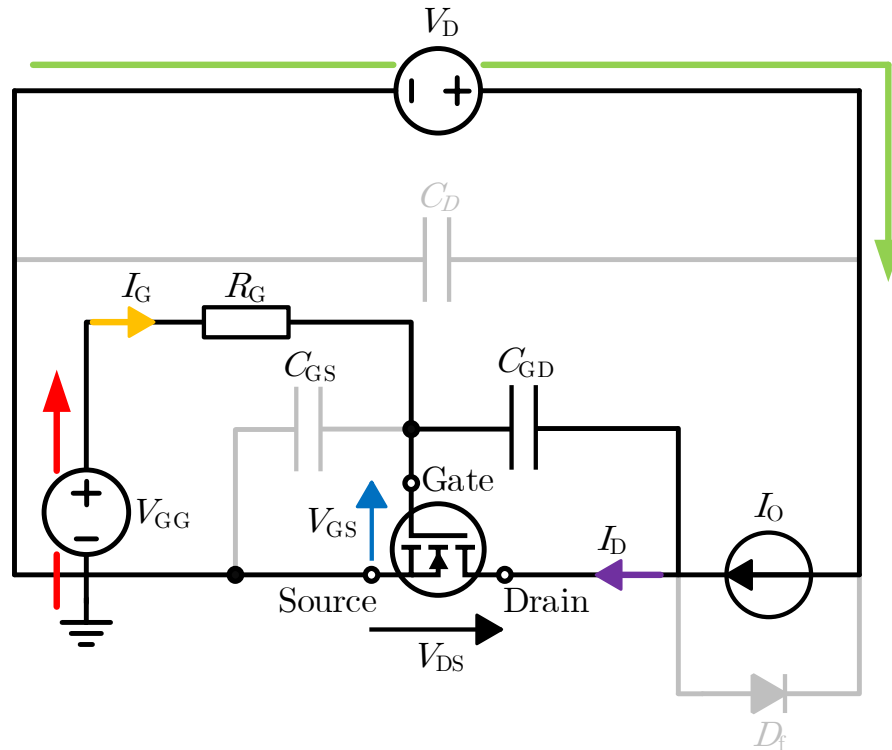
# MOSFETs – switching losses

- Threshold voltage ( $V_{GS(TH)}$ ) is reached.
- Switch starts conducting current ( $I_D$ ).



# MOSFETs – switching losses

- All  $I_O$  into the switch as  $I_D$ .
- No current through  $D_f$  and  $C_{GS}$ .
- $v_{DS}$  reduces as switch is fully turned on. This is the transition from ohmic region to active region.



# MOSFETs – switching losses

- $V_{GS}$  continues to rise towards  $V_{GG}$ .
- Switch fully on so  $V_{DS(ON)} = I_D R_{DS(ON)}$ .

